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Fourier reconstruction of signal

by

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# Abstract

This thesis addresses the problem of interpolating missing samples in seismic data. Interpolation is an important step in the seismic data processing flow, since most of the pre-processing algorithms are designed to work with regularly sampled data.

In this thesis, I will focus on the Fourier reconstruction techniques which do not require a geological model as input, but use *a priori* information of the signal to reconstruct the wavefield. This reconstruction requires that the signal be band limited.

Most presently available algorithms are slow, because of inability of fast Fourier transform (FFT) to work with irregular samples. In case of the irregularity, the Discrete Fourier transform (DFT) kernel is used for domain mapping. The computational complexity of the DFT is  $O(N^2)$  as compared to  $O(N\log N)$  for the FFT. This extra computing time can cause processing delays so seismic data processors to use simple regularization techniques which compromise accuracy for speed.

To address this problem, Non uniform Fast Fourier Transform (NFFT) is implemented, which reduces the complexity of the DFT to  $O(N\log N)$  comparable to the FFT. It has been used for seismic data regularization before using a Gaussian convolution kernel. Our newly proposed approach uses a Kaiser Bessel filter for convolution, which gives a better result.

We applied this technique to solve the problem of clipped amplitudes in Ground Penetrating Radar data. The NFFT is hybridized with the POCS (Projection on Convex Sets) method to restore clipped peaks from an acquired GPR data set.

# Acknowledgements

I would like to thank my supervisor Dr. Rob Ferguson for his excellent support throughout my program, and providing me with so many inspiring ideas. I am very grateful to Dr. Daniel Potts and Dr. Stefan Kunis for their NFFT software package. My sincere thanks go to all of my colleagues at the University of Calgary, especially to Marcus Wilson who eased the understanding of complex Mathematical analysis and to Kevin Hall who always had to solve my computer problems. A special thanks to Dr. Mostafa Naghizadeh for his guidance in understanding the concept of the Fourier reconstruction. I can't say enough thanks to my friend Ritesh Kumar Sharma who stood by my side throughout the program and also made my shift to geophysics a smooth one. My gratitude to my friend Anshita for her support, and for cooking me the best meals I ever had during college life. Above all, I want to express my most cordial gratitude to my parents for their love and support over the years.

## DEDICATIONS

To my father, Sailesh Gulati and my mother, Savita Gulati.

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# Chapter 1

## Introduction

The objective of any processing method is to refine the image of the subsurface obtained during acquisition. In exploration geophysics, our goal is to acquire, process and invert seismic wavefields to explore the potential hydrocarbons among the complex structures underneath the earth. In these methods, a seismic energy source, generally generated by an explosion, vibroseis truck or air gun travels through the earth's crust. Discontinuities in the earth's composition cause this energy to reflect and travel back to the surface. The travel time of this energy captured by an array of geophones. This gathered information is processed to improve the signal to noise ratio enhancement and finally, used to define an image of subsurface via inversion algorithms. The image produced in the end is used as a initial platform for the geophysical interpretation for exploring the geological structure, sedimentological models and process that in applied seismology can be use to delineate and exploit accumulation of hydrocarbons.

### 1.1 The need for Interpolation

Regular sampling is one of the major concerns during acquisition of seismic data. The process of acquisition prefers to record a regular finite number of spatial samples of the continuous wavefield. Regular distribution of sources and receivers leads to a better quality image. But, in the field actual sampling of seismic data is generally far from this ideal condition. In difficult terrain, due to manual error or some technical irregularity it is possible to have missing and corrupted traces in the data. Ignoring this irregularities can result in a distorted subsurface image. This irregular sampling is a major burden on

many data processing algorithms, including wave equation migration and many multiple removal methods which require regularly and densely sampled data as input.

During seismic data acquisition, the continuous wavefield is sampled as a discrete wave field based on the grid of survey. For reconstruction of the continuous wave field acquisition geometry i.e for mapping the irregular data on the regular grid, sampling rate for any axis must be equal or greater than twice dominant frequency of continuous signal (Unser, 2000). The source and receiver interval's should be decided based on the Nyquist rule (Vermeer, 1990). When this rule is neglected, interpolation comes in to play to reconstruct the data to a populated distribution of source and receivers and reproduce an approximation of the original survey (Liu and Sacchi, 2004). The quality of the reconstructed data directly affects the various steps of processing such as Migration (Spitz, 1991), AVO analysis (Sacchi and Liu, 2005), imaging (Liu and Sacchi, 2004) and noise removal (Abma and Kabir, 2005) (Soubaras, 1994).

## 1.2 Reviews of Seismic data Reconstruction Methods

Seismic data reconstruction methods are all related in the sense that they are tasked with restoring the spatial continuity of wavefield (Naghizadeh and Sacchi, 2010). Seismic data reconstruction algorithms are divided in to two categories: those based on wave equation analysis and those based on parametric analysis.

Wave equation based methods work using a regression approach, using the physics of wave propagation to reconstruct the missing samples. In geophysics, numerous approaches based on this model are proposed (Ronen, 1987; Bagaini and Spagnolini, 1999; Stolt, 2002; Trad, 2003; Fomel, 2003). Wave equation based methods require *a priori* knowledge of velocity model as input.

Parametric analysis based reconstruction methods are based on the *a priori* information from the seismic data alone (Naghizadeh and Sacchi, 2009a). Most parametric reconstruction methods are based on Fourier transformation from one domain to another (Naghizadeh and Sacchi, 2008b,c, 2009b). In the last few years excellent research is been done in this area (Liu and Sacchi, 2004; Schonewille et al., 2009). The prior assumption is either based on the stationarity of the process or based on the fact that most of the power in the power spectrum is concentrated on the lower frequencies, analysis based on this fact known as bandlimitness (Naghizadeh and Sacchi, 2010). Bandlimitness enforces only the use of certain set of frequencies (Feichtinger et al., 1995b). These algorithms perform efficiently even in situations where bandlimited assumption is not satisfied exactly (Trad, 2008).

Seismic data reconstruction is based on data mapping, generally mapping of spatial domain data to the Fourier domain. The most common bases for obtaining high resolution reconstruction techniques are the Fourier transform (Sacchi et al., 1998; Xu et al., 2005; Liu and Sacchi, 2004; Naghizadeh and Sacchi, 2007b, 2008a, 2009b, 2007a) and the Radon transform (Darche, 1990; Verschuur and Kabir, 1995). In the parabolic Radon transform, two CMP (Common Mid Point) gathers are combined to improve offset sampling and thus differences between midpoint positions are ignored (Duijndam et al., 1999). Similarly hyperbolic and linear Radon transforms (Thorson and Claerbout, 1985) as well as the parabolic Radon transform are suitable for estimating frequencies at irregular nodes, but they suffer aliasing problem due to sparse sampling (Hugonnet and Canadas, 1995), the local Radon transform and the curvelet transform (Hennenfent and Herrmann, 2006b, 2007, 2006a). Another group of signal processing interpolation methods rely on prediction error filtering techniques (Wiggins and Miller, 1972; Spitz, 1991) and (Porsani, 1999) introduced seismic trace interpolation methods using prediction filters. These methods operate frequency-space ( $f-x$ ) domain. The low frequencies in a regular spa-

tial grid are used to estimate the prediction filters needed to interpolate high frequency components. This regular spatial grid requirement restrict prediction filter methods to regular sampling.

### 1.2.1 Factor affecting Seismic Data Reconstruction

It is important to identify the factors which can affect the reconstruction technique. In my thesis, I check the effectiveness of given methods by using them on various conditions. These conditions are created by use of one or more factors given below

1. **Sampling:** The sampling operator has a significant impact on the reconstruction of seismic data. The sampling operator is a function which can be applied to regular data to model trace decimation. Regularization can be thought of as the inverse of this operator. Regular sampling function results in a regular but more sparsely sampled seismic data set and an irregular sampling function is responsible for a randomly sampled data set. The interpolation techniques work differently for the regular and irregular seismic data. Currently available techniques generally prefer regular sampling to irregular sampling. However new techniques based on the theory of compressed sensing recovered the data on the basis of superposition of a small number of basis functions.
2. **Aliasing :** The aliasing plays a major role in determining the effectiveness of an algorithm. Aliasing results in the duplication of events. Lower aliasing is easy to remove and reconstructing the missing sample. But, in case of highly aliased events, most of the algorithms fails (Naghizadeh and Sacchi, 2007b, 2008a, 2009b, 2007a). On, the basis of this many researchers advocate irregular sampling to avoid aliasing.

3. **Decimation factor:** Seismic reconstruction algorithms cannot be expected to succeed on every input. Each algorithm has its own maximum decimation factor, above which the algorithm will start fail. Thus it is important to test the limitations of each algorithm.
4. **Dimension of data :** Fourier methods can always be extended to higher dimensions. Methods which make use of all dimensions of the data, perform better. This is due to the fact that if one spatial dimension is poorly sampled, then the other dimensions algorithm can be used to reconstruct the missing samples. In this thesis algorithms are tested on just one dimension, and can be extended simply to several dimension.
5. **Event orientation:** Reconstruction algorithm produce satisfactory results when tested on seismic data with low dips. Events with high dips are always the one difficult to recover (Naghizadeh and Sacchi, 2008a, 2009b). It is important for the algorithm to be tested on steeply dipping events. For curved events, Spatial windowing is always recommend. Spatial windowing will approximate curved events as locally linear events.

### 1.2.2 Motivations

The main motivation of this thesis is to introduce Fast Fourier reconstruction techniques that can be used for seismic data reconstruction. The objective is to provide reconstruction method that can:

- be used to reconstruct regular as well as irregular seismic data.
- speed up presently available Fourier reconstruction techniques.
- handle aliased energy.

### 1.2.3 Contribution

The main contribution of this thesis are:

- The introduction of a new fast kernel for seismic data regularization.
- The hybrid approach using this kernel for declipping acquired GPR data.

### 1.2.4 Organization of this thesis

This thesis is organized as follows:

- **Chapter 2** presents the popular Adaptive weights Conjugate gradient Toeplitz (ACT) algorithm for signal construction. This algorithm is slow but accurate, and validated on several typical trace regularization situations.
- **Chapter 3** introduces the faster version of ACT, we introduced the Kaiser Bessel Non uniform Fast Fourier transform kernel (NFFT). The Kaiser Bessel NFFT kernel balances accuracy and computational cost, and we present an application of this NFFT for seismic trace interpolation. Application of the Bessel kernel for non-uniform samples is not a new algorithm, but it is an approximation scheme that can be used to calculate an approximate spectrum. In one dimension, the computational complexity of Kaiser Bessel NFFT is  $O(N \log N)$  which is a dramatic improvement from the  $O(N^2)$  complexity of the DFT use by ACT, and it is comparable to the FFT. This algorithm can be easily extended to higher dimensions. Least squares is used to refine an approximated Fourier spectrum followed by simple Inverse Fast Fourier transform (IFFT). The applicability of the proposed method is examined using synthetic seismic data.
- **Chapter 4** deals with the common problem of clipped data encountered during GPR acquisition. The Kaiser Bessel NFFT with well known POCS (Projection

on convex set) is applied. This hybrid approach is been tested on synthetic data, finally leading to some real GPR data set processing.

- **Chapter 5** concludes the thesis with summary of results.

# Chapter 2

## Band limited signal reconstruction

### 2.1 Summary

In this chapter Adaptive weights Conjugate gradient Toeplitz (ACT) algorithm for signal reconstruction is implemented. This algorithm is fast and accurate, and to test its effectiveness several typical trace regularization situations are implemented. This algorithm requires an estimate of the bandwidth as input, and overestimating the bandwidth can cause spurious high frequency noise in the reconstruction.

### 2.2 Introduction

Trace interpolation is often employed in seismic data processing. The seismic wavefield is often sampled irregularly due to economic and physical constraints, as well as technical issues. In order to exploit many efficient numerical methods to process this data, it must be projected onto a regular grid. Countless techniques exist that attempt to reconstruct seismic signals from an uneven set of samples (see (Gulati and Ferguson, 2010) for several examples). However only a few reliable methods exist to reconstruct irregularly an irregularly sampled time series without any *a priori* information (Adorf, 1995).

In this chapter, implementation of “second generation” algorithm, due to (Feichtinger et al., 1995a) to estimate the Fourier components of an irregularly sampled band limited signal, or to resample a signal onto a regular grid, using conjugate gradients on a Toeplitz matrix derived from the Fourier transform. Our goal is to assemble a toolbox of numerical methods for use and study by staff, students and sponsors of CREWES.



## 2.3 Theory

### 2.3.1 ACT Method

To derive the ACT algorithm its noted that, given an irregularly sampled signal  $s_j = s(t_j)$ , for  $j = 1, 2 \dots N$ , simple DFT of the observed samples can be computed (Vio et al., 2000),

$$S_k = \sum_{j=1}^N s_j e^{-2\pi i t_j k/N}. \quad (2.1)$$

Now, the  $S_k$  are not the true Fourier coefficients of the underlying continuous signal  $s(t)$ , which can be related to the  $s_j$  by the inverse DFT equation,

$$\begin{aligned} s_j &= s(t_j) \\ &= \frac{1}{N} \sum_{m=-M}^M \hat{S}_m e^{2\pi i t_j m/N}, \end{aligned} \quad (2.2)$$

in the case that  $s(t)$  is band limited by  $M$ . Substituting for  $s_j$  in Equation 2.1 gives us

$$\begin{aligned} S_k &= \frac{1}{N} \sum_{j=1}^N \sum_{m=-M}^M \hat{S}_m e^{2\pi i t_j m/N} e^{-2\pi i t_j k/N} \\ &= \frac{1}{N} \sum_{m=-M}^M \hat{S}_m \sum_{j=1}^N e^{2\pi i t_j (m-k)/N}. \end{aligned} \quad (2.3)$$

Writing Equation 2.3 with the true Fourier coefficient  $\hat{S}_m$  as the unknowns input and the observed Fourier coefficients  $S_k$  as the known output, the problem becomes a matrix inversion on a Toeplitz matrix, with  $N$  rows and  $2M + 1$  columns. It can be think of this as performing a discrete deconvolution on the observed Fourier coordinates, and the system can be solved for the true Fourier coefficients provided  $2M + 1 \leq N$ . Furthermore, since the matrix is Toeplitz, it can be applied to a vector in  $O n \log n$  (Feichtinger et al., 1995a), so it can expect that a conjugate gradient inversion of this matrix is fast.

This is just one of many ways to rewrite the DFT to derive a method for performing band limited signal reconstruction. Any other permutation would result in an algorithm that is technically equivalent, assuming perfect arithmetic. However in practice these

methods will have different properties and one can be more effective than another in certain situations (Vio et al., 2000).

One drawback to this method occurs when the sampling pattern is very irregular, such as when the majority of samples are concentrated in one place, which causes the reconstruction to be biased towards this area. To combat this add in a set of weights defined by the distance between a points two nearest neighbours. This will cause the densely sampled points to have lower weight in the inversion. The weights are given in equation 2.4.

$$w_j(x) = \begin{cases} \frac{1+t_2}{2} & \text{if } j = 1 \\ \frac{N-t_{j-1}}{2} & \text{if } j = N \\ \frac{t_{j+1}-t_{j-1}}{2} & \text{otherwise} \end{cases} \quad (2.4)$$

Multiplying the  $s_j$  by these weights in Equation 2.2 gives us the weighted inversion function on which the ACT method is based, given by (Feichtinger et al., 1995a),

$$S_k = \frac{1}{N} \sum_{m=-M}^M \hat{S}_m \sum_{j=1}^N w_j e^{2\pi i t_j (m-k)/N}. \quad (2.5)$$

This matrix equation is used to form the normal equations, which are solved by conjugate gradients (Shewchuk, 1994).

## 2.4 Examples

To showcase the properties of the ACT method, algorithm will test its reconstruction performance for both uniform and random decimation. For regular decimation, It can expect to note the presence of coherent noise in our reconstruction. This noise, if present, will be highly structured with strong amplitudes. For random decimation, it is expected that any error in our reconstruction will also be random, and the power will be concentrated at a few Fourier coefficients. For this algorithm, I restrict the analysis to one

dimensional signals, where the problem can be thought of as the reconstruction of a time series.

### 2.4.1 Uniform Decimation

Figure 2.1a shows a simple signal composed of two superimposed harmonics. The top panel shows the true signal, and the lower panel shows a uniform sampling of 50% of the signal. Figure 2.1b shows the discrete Fourier transform of the decimated signal. Note that the distortion of the spectrum is highly structured with high amplitude aliases. Inserting zeros into the signal to denote the missing traces results in the Fourier spectrum in Figure 2.1c. Note that the spectrum is the same but with more detail. Figure 2.1d shows the reconstructed harmonic after two iterations, and Figure 2.1e shows the Fourier spectrum. The ACT method converges linearly in relative error to the solution in the case of uniform decimation, and this simple signal is perfectly reconstructed.

The top of Figure 2.2a shows a seismic signal composed of 650 samples from a 25Hz Ricker wavelet convolved with a random reflectivity series, and the bottom panel shows the same signal with 50% of the samples set to zero. Figure 2.2b shows the reconstruction of the signal in the time domain. The reconstructed signal agrees quite well with the original signal. As with the last example, the ACT method linearly converges to the solution (Figure 2.2c), so it is very effective for uniform decimation.

### 2.4.2 Random Decimation

Figure 2.3a shows the same harmonic, but with a random selection of 40% amplitudes values set to zero. Figure 2.3b shows the reconstructed harmonic, which agrees with the original almost everywhere. Figure 2.3c shows that the relative error of the solution decays exponentially with the number of iterations. This is less desirable than the linear convergence, its been noted in the uniform decimation examples. Figures 2.4, 2.5, and

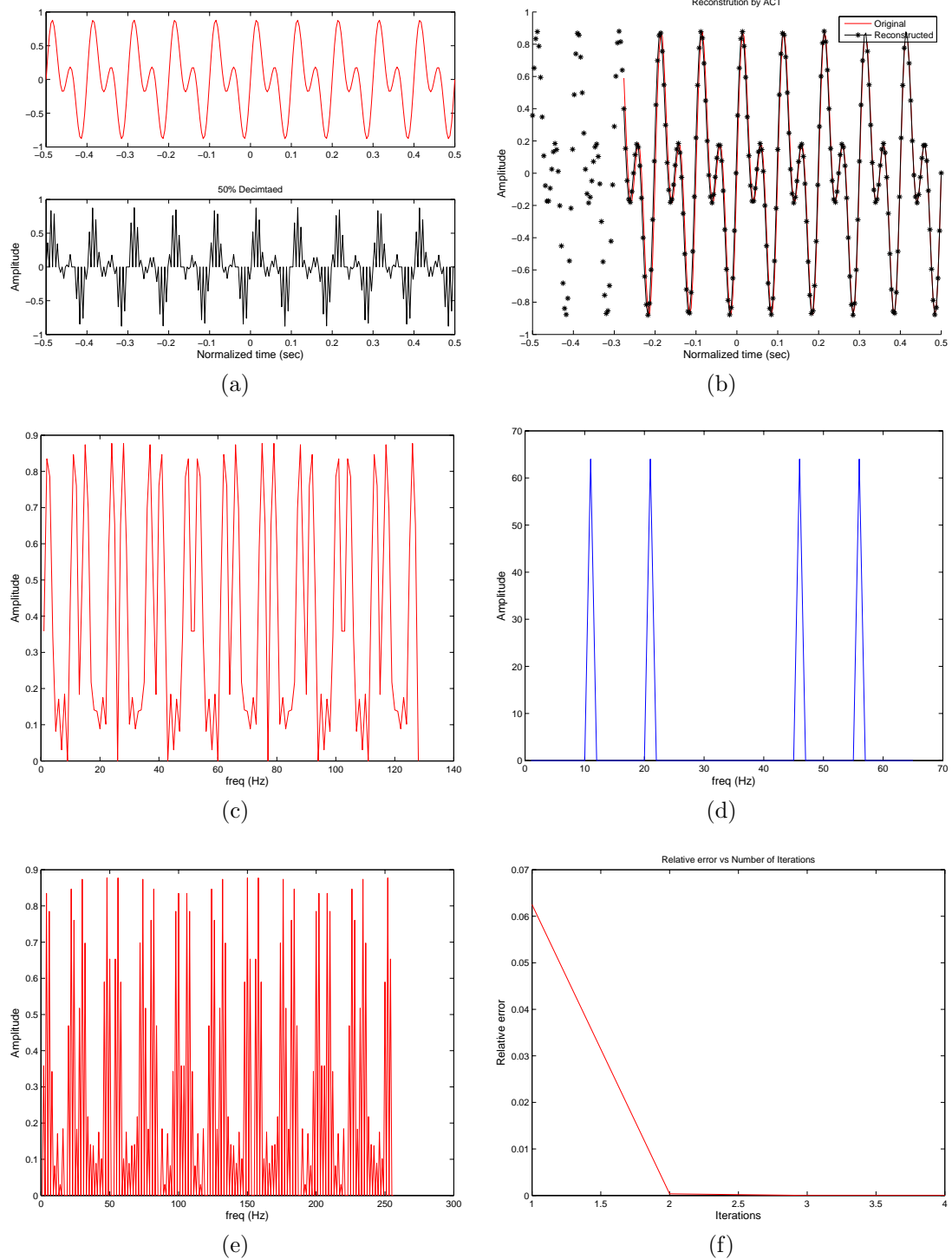
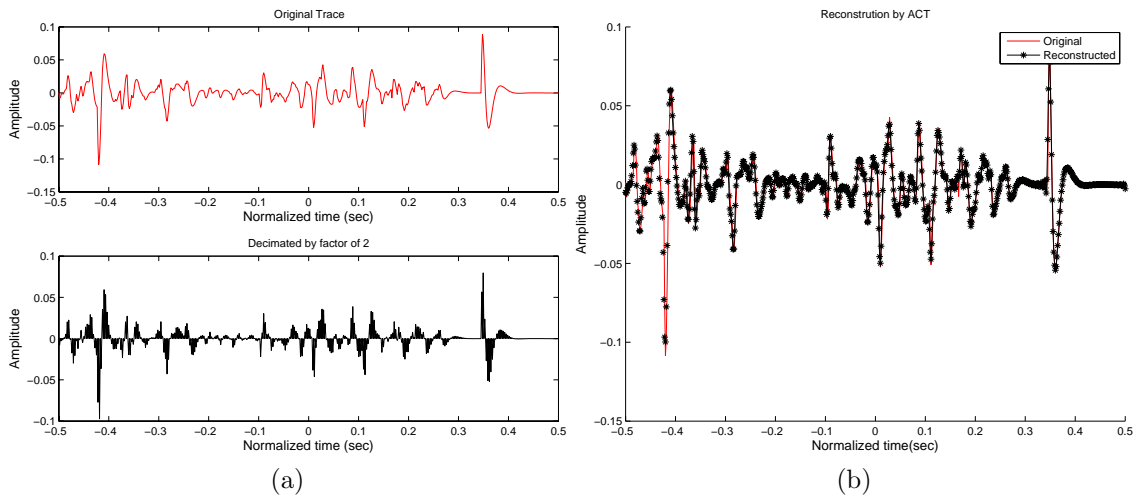
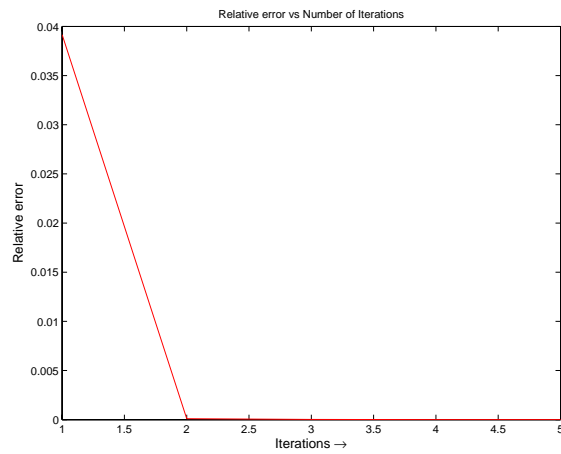


Figure 2.1: (a) A simple harmonic and a uniform 50% decimation of that harmonic. (b) The original signal and the ACT reconstruction. (c) The Fourier spectrum of the decimated signal. (d) The Fourier spectrum of the reconstructed signal. (e) The Fourier spectrum of the decimated signal with zeros in place of the unknown samples. (f) The relative error of the ACT inversion after each iteration.



(a)

(b)



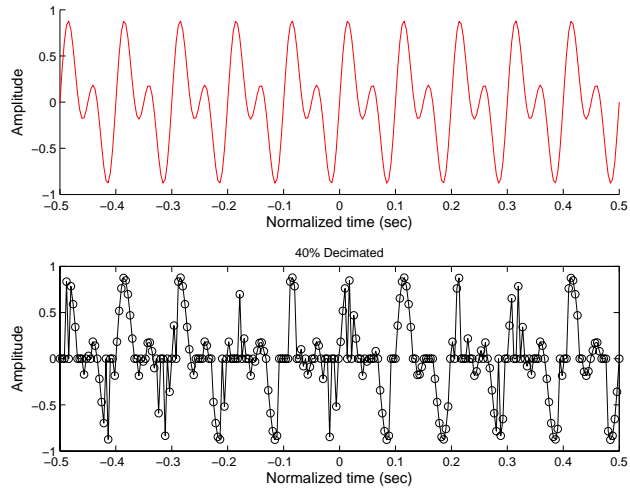
(c)

Figure 2.2: (a) A stationary seismic trace and a uniform 50% decimation of that trace. (b) The original signal and the ACT reconstruction. (c) The relative error of the ACT inversion after each iteration.

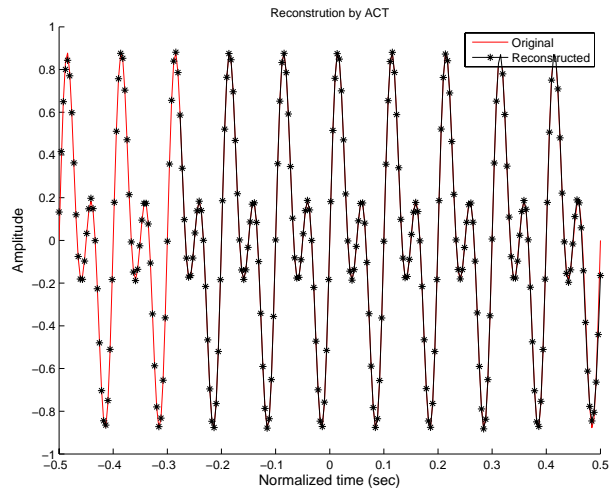
2.6 show three different random decimations of 70% of the traces. The ACT method performs well on the first trial, but breaks down on the second and third trials, resulting in significant spurious events. On observing the sampling density in Figures 2.5a and 2.6a, and the corresponding reconstructions in Figures 2.5b and 2.6b, the large anomalous peaks in the output correspond to large gaps in signal coverage. Note also that the residual error in Figure 2.5c and Figure 2.6c decays exponentially at first, but then increases in peaks in the later iterations. Figure 2.7 shows a good reconstruction for the seismic trace, randomly decimated by 30%, although the reconstruction departs from the original signal in some places. At 50% decimation this method begins to fail on the seismic trace, because the algorithm starts to map the noise to the higher frequencies (Figure 2.8).

## 2.5 Conclusion

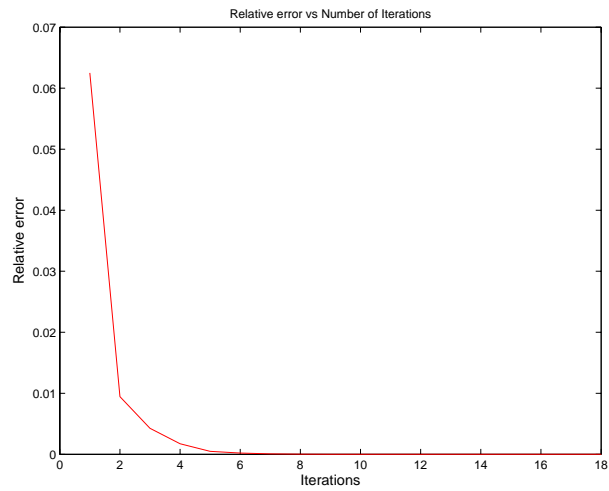
ACT method is a fast and accurate signal reconstruction method that is effective at interpolating stationary signals with up to 50% of the samples missing. The method begins to fail even on simple signals when decimation is increased to 70%, although the reconstruction can be successful if the gaps in signal coverage are not too extreme.



(a)

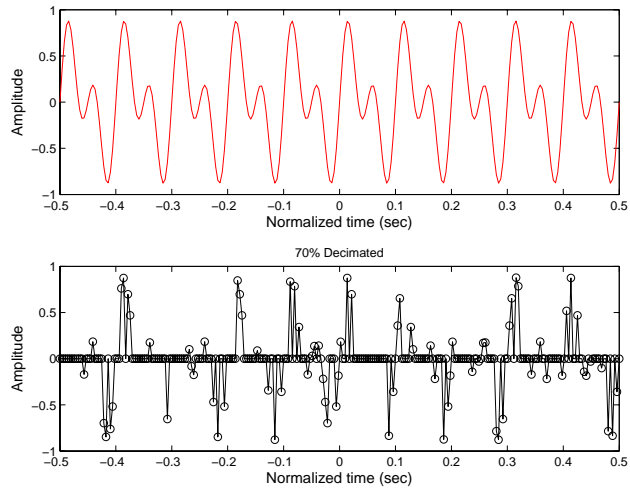


(b)

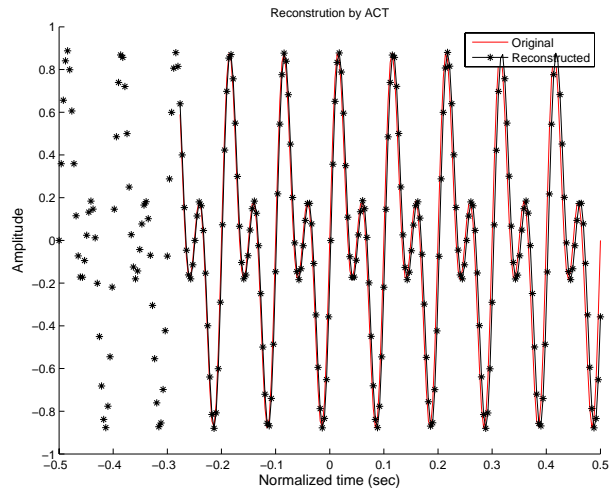


(c)

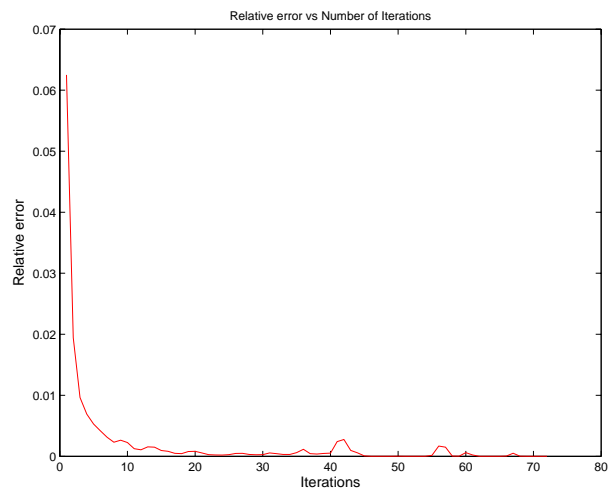
Figure 2.3: (a) A simple harmonic and a random 40% decimation of that harmonic. (b) The original signal the ACT reconstruction. (c) The relative error of the ACT inversion after each iteration.



(a)



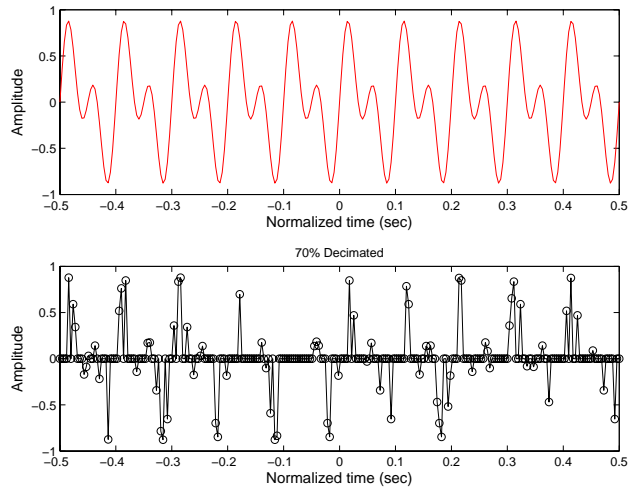
(b)



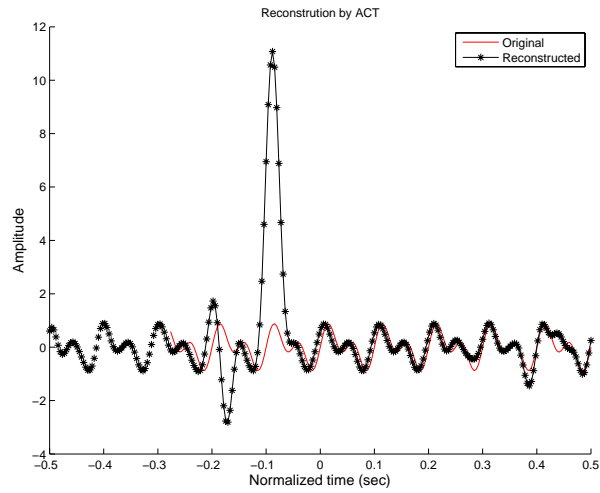
(c)

Figure 2.4: (a) A simple harmonic and a random 70% decimation of that harmonic. (b) The original signal and a successful ACT reconstruction. (c) The relative error of the ACT inversion after each iteration.

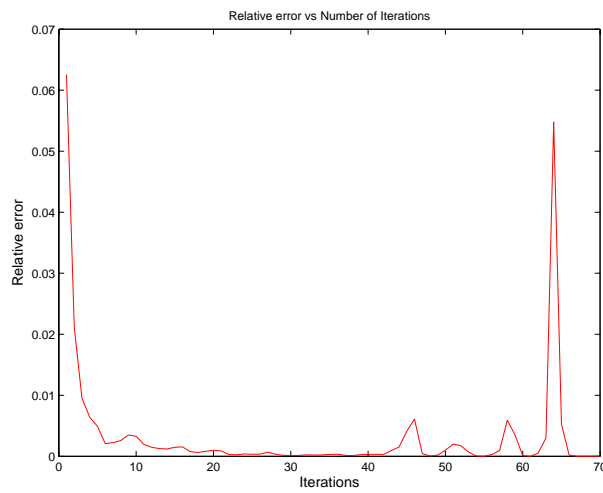




(a)



(b)



(c)

Figure 2.5: (a) A simple harmonic and a random 70% decimation of that harmonic. (b) The original signal and an unsuccessful ACT reconstruction. (c) The relative error of the ACT inversion after each iteration.

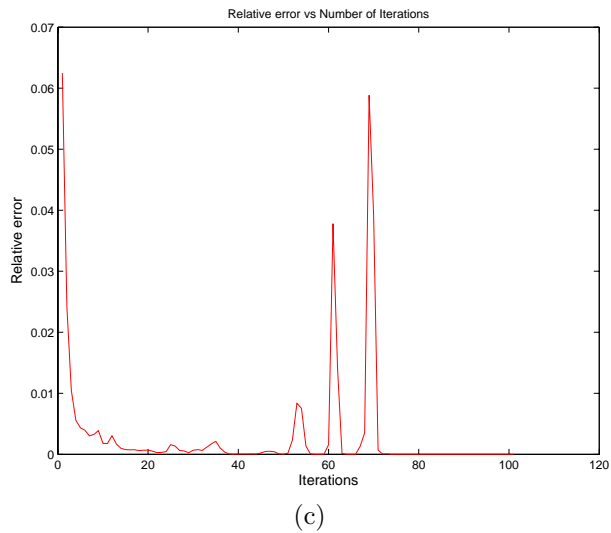
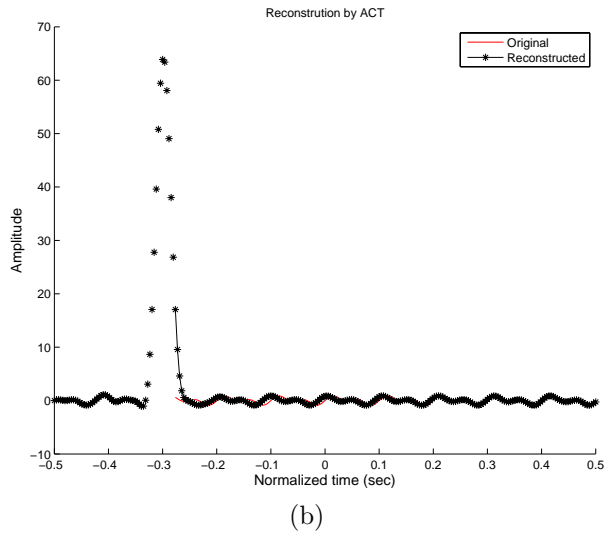
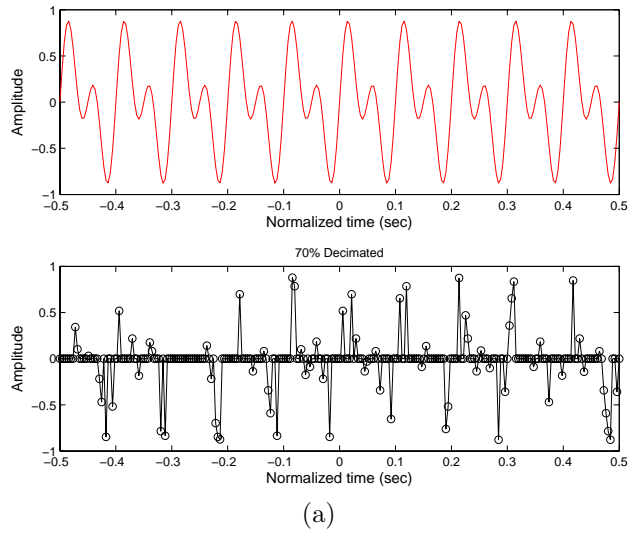
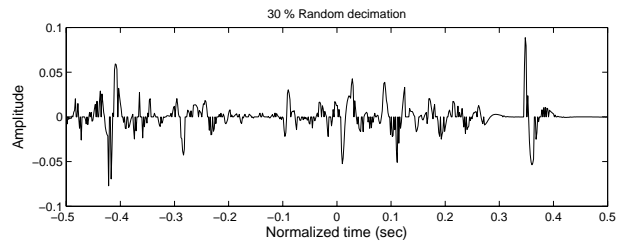
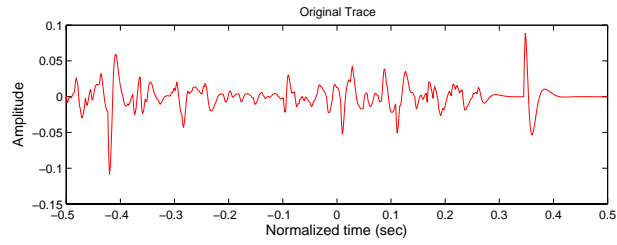
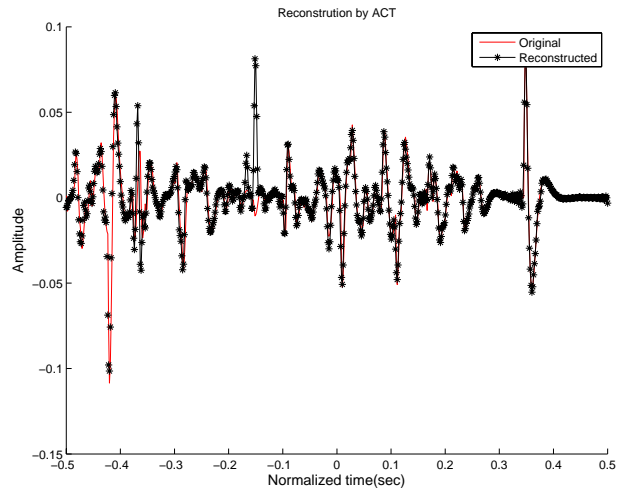


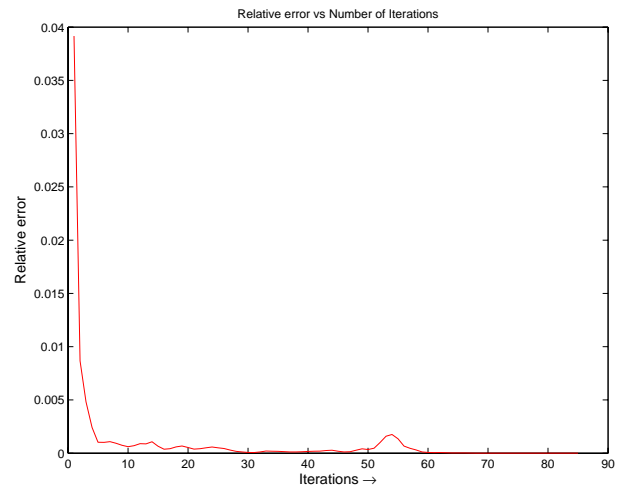
Figure 2.6: (a) A simple harmonic and a random 70% decimation of that harmonic. (b) The original signal and an unsuccessful ACT reconstruction. (c) The relative error of the ACT inversion after each iteration.



(a)

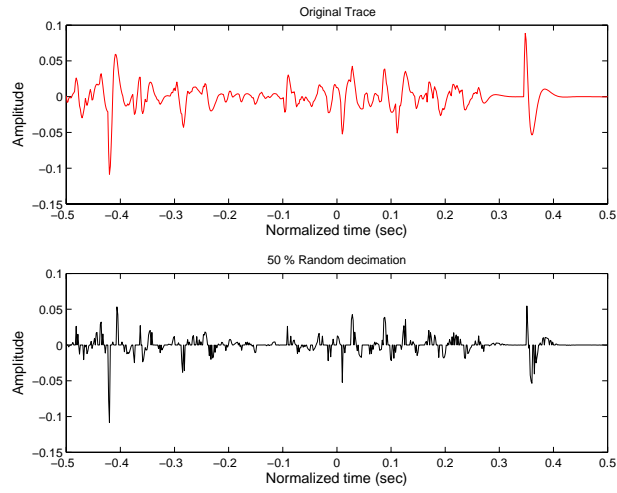


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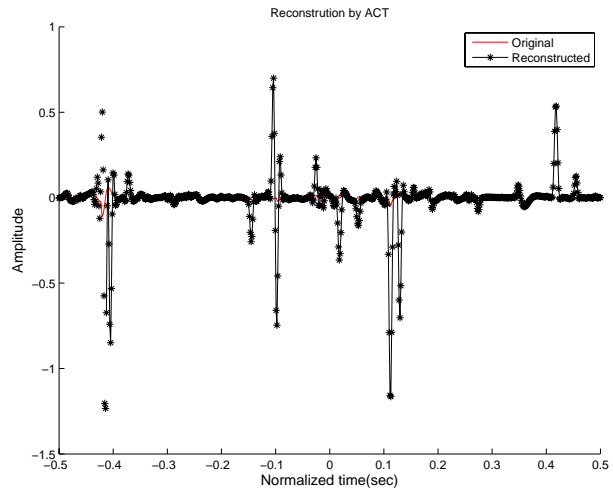


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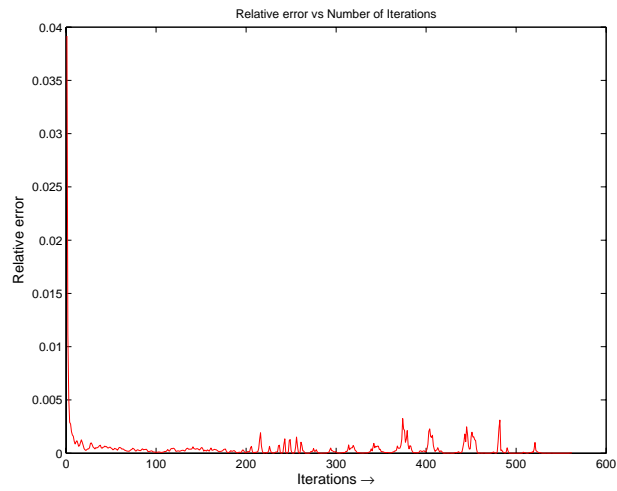
Figure 2.7: (a) A stationary seismic trace and a random 30% decimation of that trace. (b) The original signal and a successful ACT reconstruction. (c) The relative error of the ACT inversion after each iteration.



(a)



(b)



(c)

Figure 2.8: (a) A stationary seismic trace and a random 50% decimation of that trace. (b) The original signal and an unsuccessful ACT reconstruction. (c) The relative error of the ACT inversion after each iteration.

# Chapter 3

## Kaiser Bessel gridding kernel for seismic data regularization

### 3.1 Summary

There are numerous approaches which deal with the interpolation of missing samples in seismic data. The proposed Kaiser Bessel non uniform Fast Fourier transform (NFFT) kernel is new in its kind as it balances between the accuracy and computational cost. Use of Kaiser Bessel (NFFT) kernel is new for seismic data regularization.

The application of Bessel kernel for non uniform samples is not a new algorithm, but it is an approximation scheme that can be used to calculate an approximate spectrum. In one dimension, computational complexity of Kaiser Bessel NFFT is  $O(N \log N)$  which is a dramatic improvement from the  $O(N^2)$  complexity of the Discrete Fourier transform (DFT), and comparable to Fast Fourier transform (FFT). This algorithm can be easily extended to higher dimensions. Least squares is used to refine an approximated spectra followed by simple Inverse Fast Fourier transform (IFFT). The applicability of the proposed method is examined using synthetic examples.

### 3.2 Introduction

In this chapter for the seismic data reconstruction, the Kaiser Bessel window function is used. Combining the window function with Fast Fourier transform will give us the Kaiser Bessel non uniform Fourier kernel. The need of a non uniform kernel is based on the constraint that Fast Fourier transform (FFTs) needs regular spacing for its ap-

plication. Non uniform Fast Fourier transforms (NFFT) which are generalizations for the FFT are discussed by many authors in the past (Dutt and Rokhlin, 1993; Steidl, 1998; Duijndam and Schonewille, 1999; Lee and Greengard, 2006). It is important to stress that the Non uniform Fourier kernel is been used for seismic data reconstruction by Duijndam and Schonewille (1999) using B - spline and Gaussian window functions, but Kaiser Bessel window has never been tested. Proposed Kaiser Bessel based Kernel balances between the computational resources and reported to give better result than Gaussian and B- Spline window based kernels, and already been tested in Medical imaging (Knopp et al., 2007).

### 3.3 Theory

In the case of non uniform sampling, direct discretization of the forward transformation corresponding to the irregular grid at hand will be highly erroneous. A better approach will be taking the exact inverse transform from the regularly sampled domain to irregularly sampled domain and use this as a forward model in an inverse problem. The general form can be written in term of matrix vector notation as

$$Am = d \tag{3.1}$$

Where,  $A_{m \times n}$  is the forward model,  $d$  is the observation vector in time domain consist of true values and  $m$  represents Fourier components. Observation vector is irregular sampled spatial value in case of seismic data reconstruction and finally  $x$  is a unknown solution. In bandlimited approach, it will always be a over determined problem. General least square solution for such approach will be

$$m = (A^*A)^{-1}A^*d, \tag{3.2}$$

Where,  $A$  is mapping matrix from one domain to another domain and  $A^*$  represents its complex conjugate transpose.

This is basic approach for hyperbolic Radon transform and linear Radon transform by Thorson and Claerbout (1985). If desired, data estimated in the Fourier domain can be transform back to a regular grid in the spatial domain using inverse fast Fourier transform.

### 3.3.1 Discrete Fourier Transform

The general form of the forward discrete Fourier transform in case of the regular sampling can be defined as

$$\hat{P}_m = \sum_{j=0}^{N-1} P_j e^{-2\pi i n m / N} \quad (m = 0, \dots, N - 1), \quad (3.3)$$

Where,  $P_m$  are the Fourier coordinates and  $P_j$  denotes the input signal.

Assuming the regular sampling the transform can be easily inverted as,

$$\hat{P}_j = \sum_{n=0}^{N-1} \hat{P}_m e^{2\pi i n m / N}. \quad (3.4)$$

Here,  $e^{2\pi i n m / N}$  is known as the data mapping kernel. All entries of this data mapping kernel are orthogonal to each other in case of regular sampling.

The forward Discrete Fourier transform (DFT) for regularly sampled seismic data (Duijndam et al., 1999) can be written to include sample spacing as

$$\hat{P}(k_x, \omega) = \Delta x \sum_{n=0}^{N-1} P(n\Delta x, \omega) e^{-i n k \Delta x}, \quad (3.5)$$

where  $\omega$  is the temporal frequency,  $\Delta x$  is sample interval in spatial domain and  $k_x$  is the wave number. Regular sampling in spatial domain enforces periodicity.

In Equation 3.5, to avoid aliasing after the Fourier transform, it is required to keep  $\Delta x$  small. For avoiding aliasing and maintaining economics of seismic survey, it is always

better to restrict the sampling based on Shannon sampling theory (Unser, 2000). DFT is the mapping of  $N$  point signal  $(x_1, x_2, \dots, x_N)$  in to  $N$  Fourier coefficients  $X_K$ . In matrix vector form the DFT can be denoted as

$$X = DFT * x, \quad (3.6)$$

$DFT$  is a Fourier matrix that maps  $N$  dimensional vector  $x$  in to another  $N$  dimensional vector  $X$ . To transform back to the spatial domain, need  $DFT^{-1}$ , which is inverse DFT Matrix. The inverse discrete Fourier Transform is defined by

$$P(x, \omega) = \frac{\Delta k}{2\pi} \sum_{m=-M}^{m=M} \hat{P}(m\Delta k, \omega) e^{-im\Delta kx}, \quad (3.7)$$

where  $\Delta k$  is the sampling interval in Fourier domain, where  $N = 2M + 1$  and  $\Delta k = \frac{2\pi}{N\Delta x}$ .

The matrix vector form of equation 3.7 is

$$x = DFT^H * X, \quad (3.8)$$

where  $DFT^H$  is the Hermitian adjoint of DFT, Since sampling is regular,  $DFT_{N \times N}$  is orthogonal, which implies

$$DFT^H * DFT = NI_N, \quad (3.9)$$

where  $I_N$  is  $N$  dimensional identity matrix. Equation 3.9 shows that DFT is an orthogonal transformation, and that the inverse is computed using a Hermitian operator. The cost of inverting  $N \times N$  Hermitian operator is  $O(N^2)$  instead of  $O(N^3)$ . Cost is further diminished to  $O(N \log N)$  using the fast Fourier transform (FFT) instead of matrix vector multiplication. However, FFT can't be applied in the case of irregular sampling

$$DFT^H * DFT \neq NI_N, \quad (3.10)$$



Equation 3.10 shows that when sampling is irregular, it is not simple to invert the DFT matrix, since columns of the DFT matrix is no longer orthogonal. The approximation converging closest to DFT for irregular sampling is the weighted Fourier Transform (DFT).

$$P(m\Delta k, \omega) = \sum_{n=0}^{N-1} P(x_n, \omega) e^{jm\Delta k x_n} \Delta x_n, \quad (3.11)$$

where  $\Delta k$  is the regular sample interval in Fourier domain.  $x_n$  represents the positions of the irregular nodes, and  $\Delta x_n$  is the weighting factor which depends upon the distance between the samples in spatial domain.

$$\Delta x_n = \frac{x_{n+1} - x_{n-1}}{2}, \quad n = 0, \dots, N - 1. \quad (3.12)$$

The DFT in Equation 3.11, however, is not a unitary transformation, as it fails the dot product test (i.e the dot product of two vectors before the transformation should be equal to dot product after the transformation). For this reason, it is not possible to reconstruct the original domain by a simple inverse FFT (IFFT).

Feichtinger et al. (1995b) suggests an approach to handle the irregular grid problem by putting a band limitation restrain on the data. If  $\Delta k$  is the sampling interval in Fourier domain than the data is band limited to between  $[-M\Delta k, M\Delta k]$ . Accordingly, equation 3.4 for  $N$  irregular samples  $(x_0, x_1 \dots, x_{N-1})$  can be denoted in matrix vector notation as

$$y = A\hat{p}, \quad (3.13)$$

where,

$$y_n = P(x_n, \omega). \quad (3.14)$$

Where,  $y_n$  represents the values on the non uniform grid. Also,

$$\hat{p}_m = \hat{P}(m\Delta k, \omega), \quad (3.15)$$

$$A_{nm} = \frac{\Delta k}{2\pi} e^{-jm\Delta k x_n}. \quad (3.16)$$

Where,  $\hat{p}_m$  is the solution for the linear least square problem and  $A_{nm}$  is the data mapping kernel. However, real data is never band limited; there will always be some spatial frequencies above the restricted bandwidth. It can be treated as noise in the forward model and can be included in equation 4.1 as

$$y = A\hat{p} + \text{Noise}, \quad (3.17)$$

Further,  $\hat{p}$  can be estimated by

$$\hat{p} = (A^H W A + k^2 I)^{-1} A^H W y, \quad (3.18)$$

where  $W$  is a weight matrix,  $k$  is the stabilization factor and  $A^H$  is the complex conjugate transpose of  $A$ . From equation's 4.2, 4.3, and 4.4, the last term of equation 3.18 can be written as

$$A^H W y = \frac{\Delta k}{2\pi} \sum_{n=0}^{N-1} P(x_n, \omega) e^{jm\Delta k x_n} W_{nn}, \quad (3.19)$$

where  $W_{nn} = \Delta x_n$ . Here, except for constant  $\frac{\Delta k}{2\pi}$ , equation 3.19 is equivalent to equation 3.11, which represents weighted DFT. Estimated Fourier spectrum  $\hat{p}$  can be transformed back to the spatial domain by direct inverse transform. The DFT is a major computational task for the forward transform, as computational complexity of the DFT is  $O(N^2)$ . Many inversion schemes that are use in data processing (Sacchi et al., 1998; Sacchi and Ulrych, 1996) rely on the solution of normal equations on the right hand side of which is DFT.

Proposed kaiser Bessel kernel is a solution that can replace slow DFT with faster algorithm. Fast algorithm will make many algorithms where DFT is used as practical for industry.

### 3.4 Methodology

Methodology is divided into two categories forward problem and inverse problem. Both is calculated using NFFT Kaiser Bessel kernel. Methodology can be divided in following steps

1.  $A^H W y = b$ , calculates direct forward transform using NFFT kernel.
2.  $U = (A^H W A + k^2 I)$ , computation of deconvolution operator, using NFFT and Adjoint NFFT kernels.
3.  $U \tilde{p} = b$ , calculates least squares system for  $\tilde{p}$ .
4.  $y_1 = IFFT(\tilde{p})$  calculates direct backward transform on regular grid using Fast IFFT.

#### 3.4.1 Forward problem

The Non uniform Fast Fourier gridding algorithm can be numerically expressed in following steps: gridding, FFT, deconvolution. The gridding is obtained by convolution of the sampled signal values with a convolution function followed by re-sampling onto a cartesian grid. Convolution with Kaiser Bessel  $kb(x)$  is carried out to make the signal approximately band-limited according to

$$p_g(m) = kb(x) * p(x), \quad (3.20)$$

where  $p_g(m)$  is the result of spatial convolution. Equation 3.20 can be written as multiplication in the Fourier domain as

$$P_g(m) = KB(m) \times P(m), \quad (3.21)$$

where  $P_g(m)$  is the Fourier spectrum of  $p_g(m)$  in Fourier domain. For efficiency, Kaiser Bessel need to be truncated, thus generating  $n$  samples for  $p_g(m)$  where

$$n = -int\left(\frac{q+1}{2}\right) + 1, \dots, N + int\left(\frac{q+1}{2}\right) - 1, \quad (3.22)$$

and where  $int(x)$  truncates to the largest integer smaller than  $x$  for  $x \geq 0$ . The algorithm is initialized at  $p_{\tilde{g}}(n) = 0$ , where subscript  $\tilde{g}$  indicates an application of Kaiser Bessel filter and keep updating by summation of the  $N$  shifted filters. This summation of  $N$  shifted filter can be given by

$$p_{\tilde{g}}(n) \leftarrow p_{\tilde{g}}(n) + \Delta x p_n kb(n\Delta x - x_n) \quad (3.23)$$

Equation 3.23 represents mapping of the irregular samples on to a regular grid. The sampling  $p_{\tilde{g}}(n) = \Delta x p_1(n\Delta x)$  is similar to equation 3.24 in Fourier domain which can be written as

$$P_g(m) = \sum_{I \in \mathbb{Z}} P(m + IN) KB(m + IN) \quad (3.24)$$

When  $P_g(m)$  is broadband, aliasing will occur when  $KB(m + IN) \neq 0$  for any  $I \neq 0$ . It is suggested (Duijndam and Schonwille, 1999) that to remove the aliasing, there is requirement of making the signal periodic.

$$p_{\tilde{g}}(n) = \sum_{I=-\infty}^{\infty} p_{\tilde{g}}(n + IN), \quad n = 0, 1, 2, \dots, N - 1, \quad (3.25)$$

where  $p_{\tilde{g}}(n + lN) = 0$  is outside the interval given by equation 3.22. Convolution of the signal followed by the discrete transform can be represented by

$$P_g(m)_{FFT} = \sum_{n=0}^{N-1} p_{\tilde{g}}(n) e^{j2\pi nm/N}, \quad m = \frac{N}{2}, \dots, \frac{N}{2} - 1 \quad (3.26)$$

where  $P_g(m)_{FFT}$  is the spectrum obtained using the FFT. Finally correction for convolution is carried out by deconvolution in Fourier domain according to

$$P(m) = \frac{P_g(m)_{FFT}}{KB(m)}, \quad (3.27)$$

where  $P(m)$  is the approximate spectrum and  $KB(m)$  is Fourier domain representation of Bessel filter.

### 3.4.2 Window function

NFFT algorithms are based on convolution of sampled signal with a band limiting filter, and several different names are indicated in the literature. Jackson et al. (1991) discuss these algorithms in terms of image processing and refer to them as gridding algorithms. Beylkin et al. (1991) proposes a similar irregular Fourier transform algorithm where convolution with B-spline is carried out to make the signal approximately band limited. Jackson et al. (1991) discussed several forms of filters which can be used, and a truncated Gauss filter is introduced by Dutt and Rokhlin (1993).

Most of the recent development in these algorithms deals with optimization of above windows functions, but still the Kaiser Bessel window function gives the best result (Knopp et al., 2007).

#### Kaiser Bessel window function

Prolate Spheroidal wave function (PSWF) has finite time support and maximum concentration of energy within a given bandwidth. The PSWF is the eigenfunction having the largest eigenvalue of the operation of repeatedly low-pass filtering a function and band-limiting it. It is difficult to compute but the Kaiser Bessel function is a close approximation of the PSWF. For a given filter  $q\Delta x$  and bandwidth B, the least amount of energy outside desired passband i:e minimization of

$$\frac{\int_{|m|>B} |\tilde{g}(m)|^2 dm}{\int_{-\infty}^{\infty} |\tilde{g}(m)|^2 dm}. \quad (3.28)$$

Kaiser Bessel function can be represented as (Knopp et al., 2007)

$$g(x) = \frac{1}{q\Delta x} I_0 \beta \sqrt{1 - \frac{2x}{q\Delta x}}^2 \quad \frac{-q\Delta x}{2} \leq x \leq \frac{q\Delta x}{2}, \quad (3.29)$$

Where  $I_0$  is the zeroth order modified Bessel function of its first kind. In Frequency domain, its Fourier transform is used for deconvolution purpose. Fourier domain repre-

sensation of Kaiser Bessel function can be written as

$$\tilde{g}(m) = \frac{\sin(\sqrt{\pi^2(q(\Delta x)^2 m^2 - \beta^2)})}{\sqrt{\pi^2(q(\Delta x)^2 m^2 - \beta^2)}}. \quad (3.30)$$

Figure 3.1a represents Kaiser Bessel window for various value of  $\beta$  in spatial domain.  $\beta$  is the parameter for Kaiser window, which gives control over trade off between main lobes width and side lobes level. Large  $\beta$  gives wider main lobe but lower side lobes as shown in Figures 3.1a and 3.1b. For maximum frequency resolution, always narrowest main lobe is preferred. Jackson et al. (1991) carried out detailed analysis of the various convolution functions leading to the approximation for the prolate spheroidal function. Different value of  $\beta$  is suggested for 3.30 in (Jackson et al., 1991). For all calculation purpose value of  $q$  is taken as 6 and value of  $\beta = 2$ . These values are taken as a optimum by (Knopp et al., 2007).

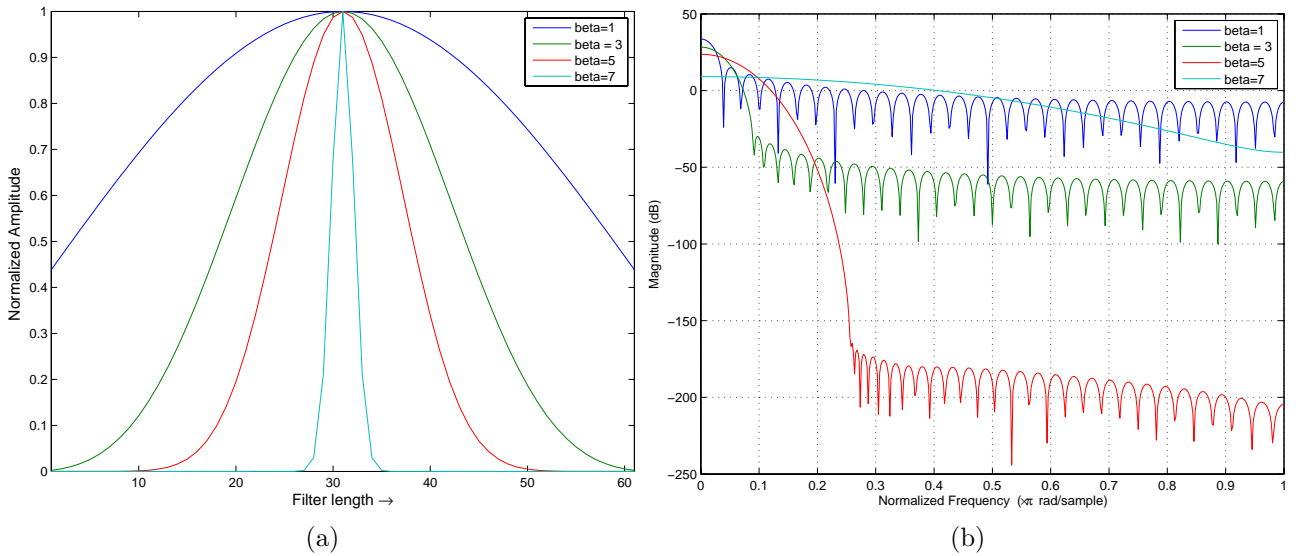


Figure 3.1: Kaiser Bessel filter. a) Kaiser window in spatial domain. b) Kaiser window in Fourier domain.

### 3.4.3 Comparision

To assess accuracy and computational complexity in signal reconstruction, comparisons are made with the traditional gaussian method in the framework of NFFT. Figures 3.2 and 3.3 show a comparative study of the above two algorithms. I tested the performance of each algorithm with a very limited dataset which require maximum 100 iteration for converging to solution. The test is performed on signal composed of random number of harmonics with 512 samples with sampling rate of .01ms. Decimation is performed with a 5% increment which decimate the signal beginning from 5% to 85%. At low decimation rate both Gaussian and Kaiser Bessel converge to the solution at faster rate with minimum number of iterations. The RMS error during the initial stages of iteration between 5% to 25% decimation did not show much difference for the two algorithms. After 25% of decimation, the Gaussian starts giving high error values for first few iteration when compared with Kaiser Bessel. Figure 3.2 shows that solution is converging to a unstable solution with higher error values where as Kaiser Bessel is giving us a stable solution with low error values. At the higher decimation rate Kaiser Bessel algorithm outperforms the Gaussian by converging smoothly at 50 % to 75 % decimation with low error values for number of iterations lies between 20 and 40. On the other hand, Gaussian starts failing giving higher error peaks while converging with higher number of iterations between 60 and 80 as illustrated on the left side of figure 3.3.

In summary, Gaussian is not suitable for higher decimation rate when compared to Kaiser Bessel. On the other hand, Kaiser Bessel performs better, in terms of accuracy and stability at higher decimation. This comparison leads to a decision of using Kaiser Bessel filter as the main kernel for implementation of NFFT for regularization problem.

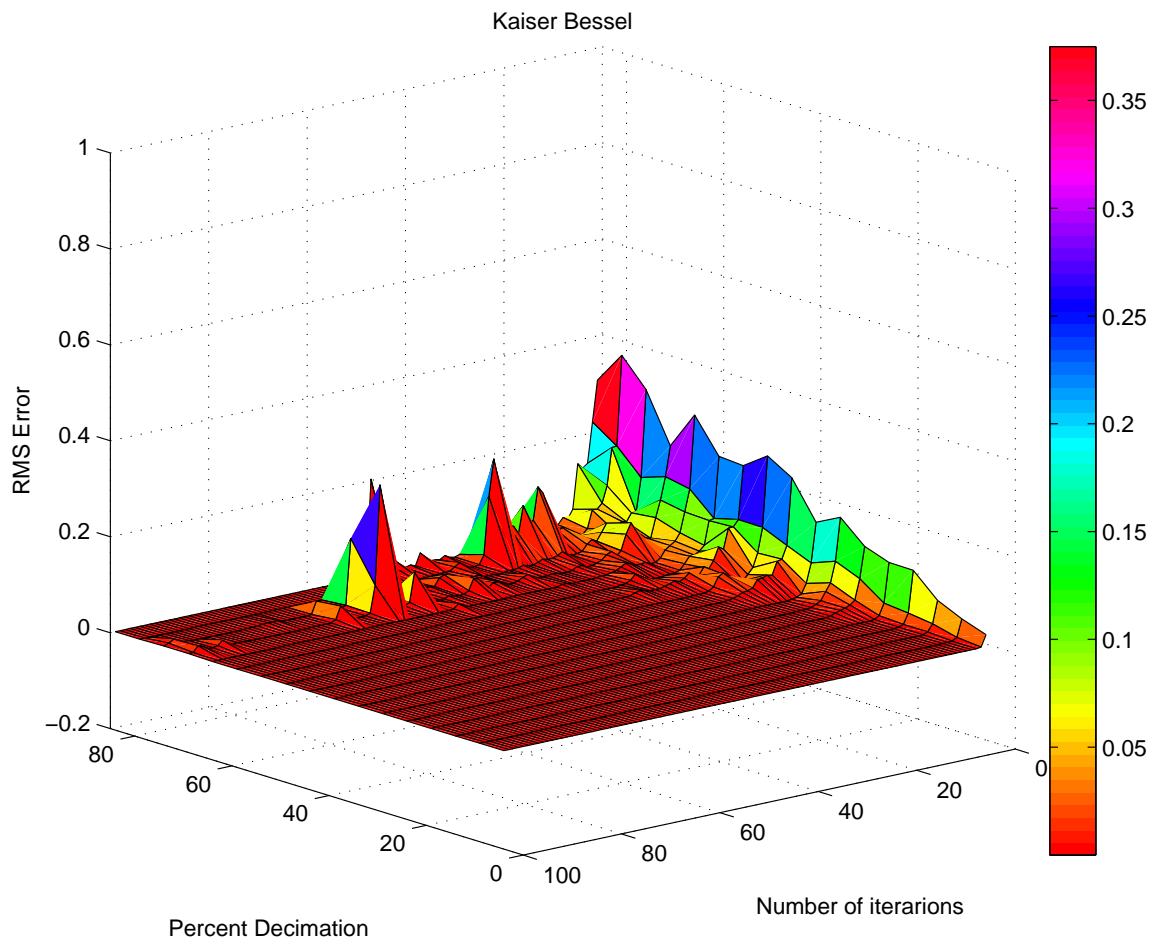


Figure 3.2: RMS error vs number of iterations and percent decimation for Gaussian window



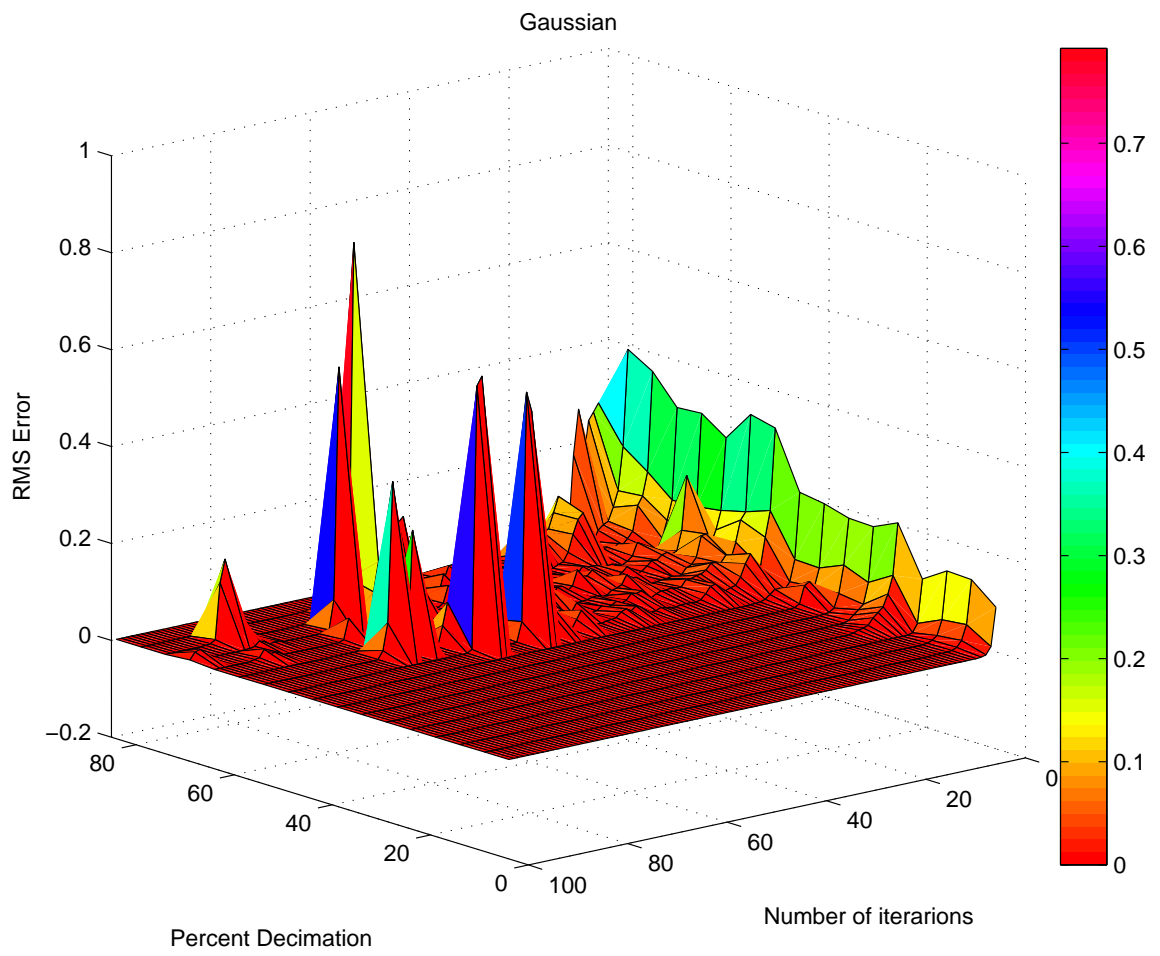


Figure 3.3: RMS error vs number of iterations and percent decimation for Kaiser Bessel window

### 3.4.4 Inversion

In general underdetermined linear system will solve in this problem, so the solution can only be approximated up to a residual of the form

$$r = y - A\tilde{p}. \quad (3.31)$$

In order to compensate for the missing samples it is important to incorporate a weight function  $W$ ,  $W > 0$  and the problem becomes a

$$\text{argmin} \|y - A\tilde{p}\|_W^2 = \sum_{j=0}^{M-1} w_j |y_j - f(x_j)|^2 \rightarrow \text{min}, \quad (3.32)$$

where  $W = \text{diag}(w_j)_{j=0, \dots, M-1}$

## 3.5 Efficiency

The problem of regularization in the least squares NFFT framework is divided in two categories: forward method and inversions. The direct forward transform is been computed using NFFT which is  $A^H$  and for the inversion purpose operator  $U = (A^H W A + k^2 I)$  is computed. For computing the inversion operator, the forward Fourier kernel  $A^H$  and its adjoint  $A$  is already computed using NFFT. It has already been that NFFT give a computational advantage over DFT. Further more iterative solution of 3.32 has been analysis in detail in large number of papers (Feichtinger et al., 1995b). The adaptive weights conjugate gradient Toeplitz method (ACT) applies the conjugate gradient method to the weighted normal equation which can be written as

$$A^H W A \tilde{p} = A^H W y. \quad (3.33)$$

## 3.6 Synthetic Tests

### 3.6.1 Synthetic 1D examples

Purpose of any reconstruction algorithm can only be solved if it is tested as general algorithm. Its important stress that not all the methods are capable of dealing with regular as well as irregular sampling. In fact, most of the parametric signal reconstruction technique fails to deal with irregular sampling (Naghizadeh and Sacchi, 2008b,c, 2009b; Hennenfent and Herrmann, 2008, 2007).

Figure 3.4 demonstrates effect of the sampling on seismic data. Synthetic hyperbolic events (Figure 3.4a) and its Fourier domain representation (Figure 3.4b). In case of regular decimation (Figure 3.4d), strong coherent noise (Figure 3.4d) will be created due to acquisition. Noise is highly structured with strong amplitudes. Most of the regular interpolation techniques is based on the idea of using non aliased low frequency and de-alias higher frequency. (Abma and Kabir, 2005) pointed out that most interpolation method based on regular sampling whereas irregular sampling generate weak noise. In irregular sampling (Figure 3.4e), power is focused at few Fourier coefficients and noise is spread whole transform domain (Figure 3.4f). Sparser the signal, straightforward will be the reconstruction.

For examining the performance of Kaiser Bessel NFFT algorithm with various sampling operators, created a simple sin signal in Figure 3.5 as well as another signal in Figure 3.6 which is composed of two harmonics. Detailed analysis with varying gaps, extrapolation, random sampling and uniform sampling is carried out.

For 1 dimension examples will take case of simple sinusoidal with 256 samples, at sampling rate of 10ms. Top panel will show the decimated spatial domain and panel below it is reconstructed missing samples. Figure 4.1a shows the 30 % randomly decimated signal and reconstructed sinusoidal. Even with 50% randomly decimation in Figure

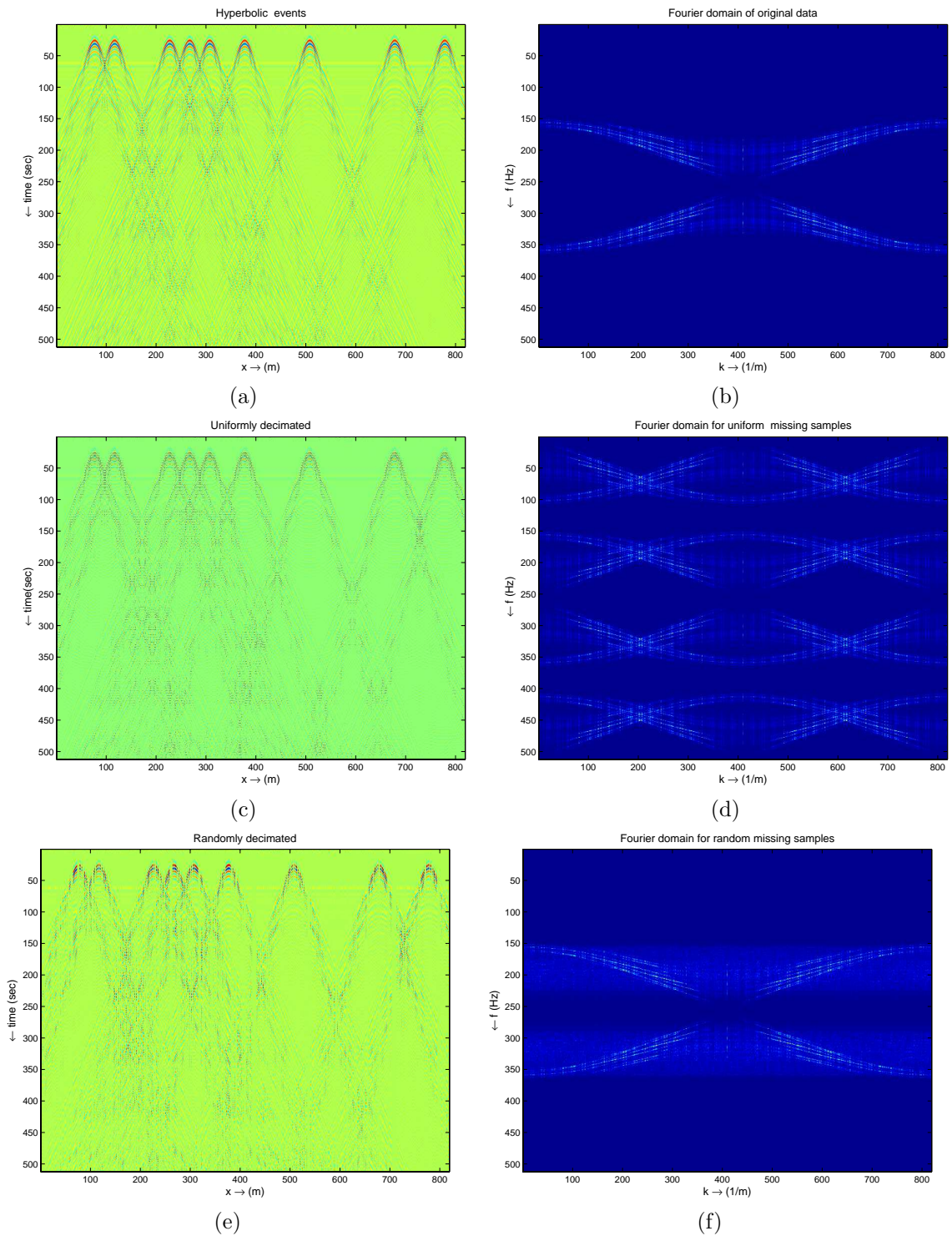


Figure 3.4: Effect of sampling on seismic data. a) Hyperbolic events in spatial domain. b) Fourier domain for hyperbolic events. c) Uniform decimation for hyperbolic events. d) Fourier domain for uniformly missing samples. e) Random decimation for hyperbolic events. f) Fourier domain for randomly missing samples.

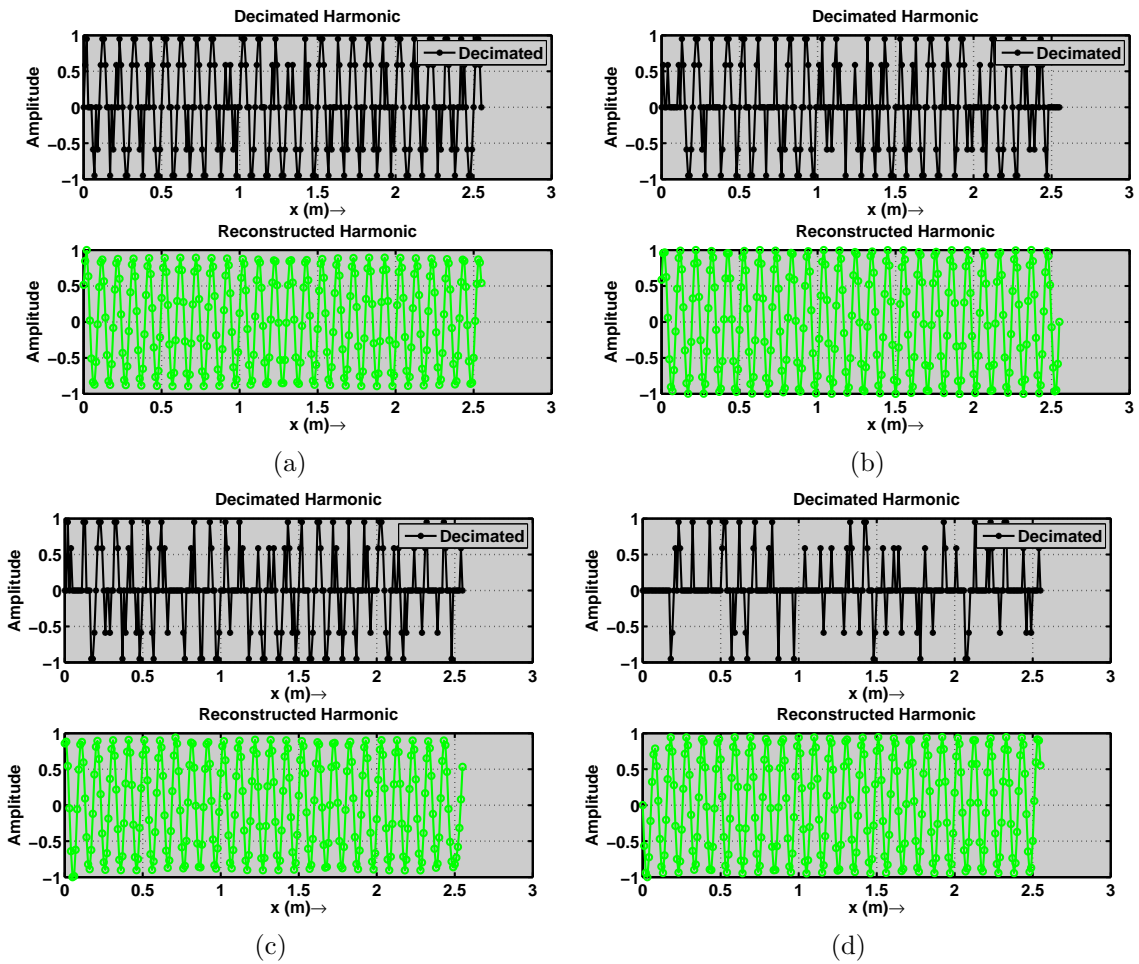


Figure 3.5: Reconstruction for Harmonics. a) Harmonics with 30 % decimation. b) Harmonics with 50 % decimation. c) Harmonics with 60 % decimation. d) Harmonics with 80 % decimation.

4.1b algorithm seems to do pretty well. On implementing high decimation sampling functions of 60% in Figure 3.5c results are good, all missing samples have been successfully reconstructed. On going further decimation in Figure 3.5d due to lost of the Fourier coefficients it is not able to reconstruct the same amplitude back, except at one point where it is missing most of the samples.

### 3.6.2 Gaps

In the previous test, it has been observed that algorithm fails some time when more number of Fourier coefficients are missing from a single location. This behaviour is further tested in gap test. In this different gaps will be created by taking more number of Fourier coefficients from a single location. Algorithm is tested for all size of gaps. Input signal composed of two harmonics , with sampling interval of 10ms for 256 samples. In case of small gaps in Figure 3.6b, reconstruction is perfect. Even in the presence of large gaps in Figure 3.6d, algorithm works effectively.

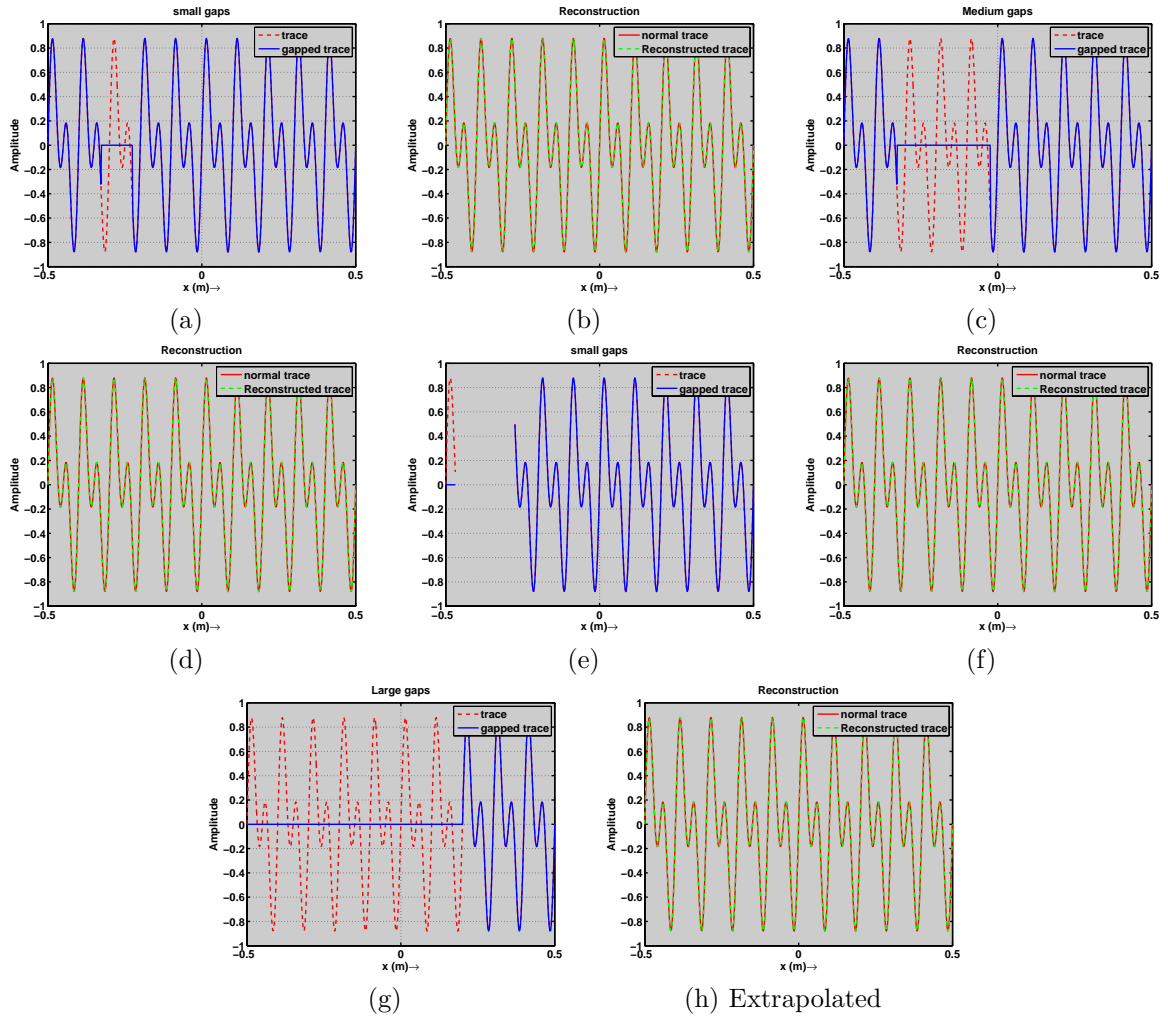


Figure 3.6: Reconstruction and extrapolation of gaps. a) Small size gaps. b) Reconstructed small gapped harmonic. c) Medium size gaps. d) Reconstructed medium gapped harmonic. e) Small side gaps. f) Extrapolated small gaps. g) Big side gaps. h) Extrapolated big gaps.

### 3.6.3 Extrapolation

Extrapolation test is done for the reconstruction algorithm, purpose of algorithm is to extrapolate the missing samples. Extrapolation is been tested on combination of two harmonics for two categories, small gaps and large gaps in Figures 3.6e and 3.6g. Reconstructed extrapolated harmonics can be seen in Figures 3.6f and 3.6h. Algorithms can easily handle the stationary harmonics with large gaps. Algorithm can also be applied on simple non stationary harmonics when taken small windows, and events are assumed to be stationary.

## 3.7 Synthetic 2D examples

In case of 2D data reconstruction, the Fourier reconstruction is iterative on each frequency slice in FK domain. NFFT least square will be applied on each frequency slice, with iteratively moving to next slice. In Figure 3.7, there are three seismic events with different dips and amplitudes. The seismic wavelet is ricker wavelet with peak frequency of 50 Hz. Sampling rate for seismic data acquisition is 4ms. Figure 3.7 is an original synthetic section. Figure 3.8 represents Fourier domain representation of original section. Before testing algorithm for heavy decimation operators, its been tested for 10% random decimation in Figure 3.8. NFFT least squares works perfectly in Figures 3.7 and 3.8 for the small random decimation.

### 3.7.1 Randomly decimated dipping Events

Random sampling in the spatial domain (Figure 3.9a) can result in low amplitudes artifact like in Figure (3.9c) along with the original Fourier events. The artifacts are the resultant of random sampling operator which is 50% resultant due to decimation in original data in Figure 3.9a. Reconstructed data in Figure 3.9b in case of 50% random decimation is



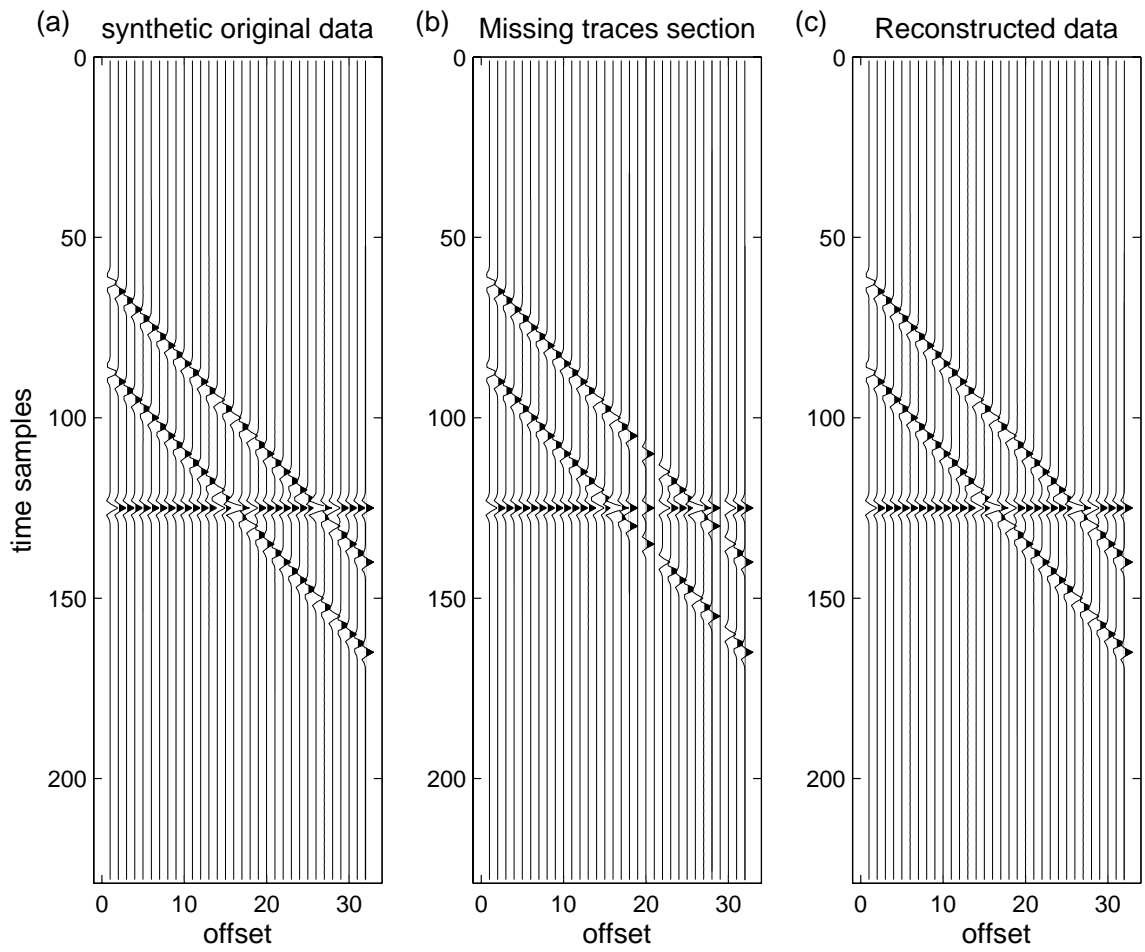


Figure 3.7: Synthetic seismic data. a) Synthetic original data. b) Missing traces section with 10 % decimation. c) Reconstructed traces for synthetic data.

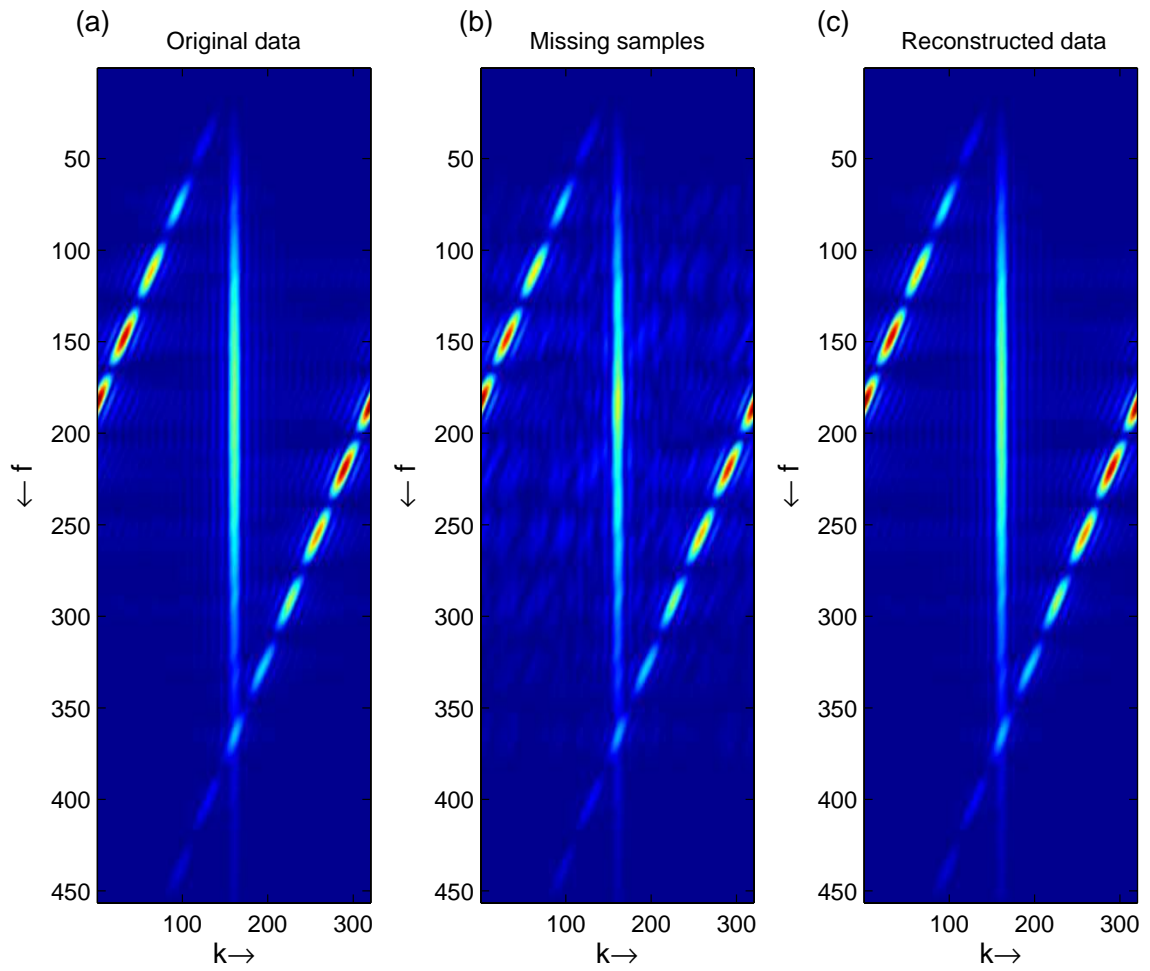


Figure 3.8: Fourier domain representation. a) Fourier domain for original data. b) Fourier domain for missing traces with 10 % decimation. c) Fourier domain for reconstructed data. of original event with 10 % decimated data and reconstructed data.

as good as original. Figure 3.9b proves that algorithm works for the seismic section with half of the missing samples. Even Fourier domain in Figure 3.9c shows all the energy concentrated on the dipping events, with no energy getting dissipated.

Further moving to higher decimation of 80% in 3.9e low amplitudes artifacts are more dominant. Along with the dominant artifacts, aliasing for the dips can be seen in Figure 3.9g. Noisy artifacts are observed in Figure 3.9f as compared to Figure 3.9b, it is because of the big gap in Figure 3.9e. It was seen before that algorithm works for big gaps in case of simple harmonics in Figures (3.6b, 3.6d), but it is effective even in case of linear dipping events. It is important to test when algorithm fails for knowing its limitation. Therefore final data is tested using random sampling operators of 80% decimations in Figure 3.9f. Figure 3.9e shows 80% decimated data, with its Fourier domain in Figure 3.9g. It should be noticed that Fourier domain of 80% decimation in Figure 3.9g has more aliased events than with 50% decimation, it is again due to the presence of more gaps in the decimated section in Figure 3.9e as comparison to 50% decimation in Figure 3.9a. Algorithm started to fails with 80% decimation as seen in reconstructed section in Figures 3.9f, there are low amplitudes artifacts in the recovered Fourier domain (Figure 3.9h) as well. Events in recovered section are still well defined (Figure 3.9f) but with the high amplitude noise in the section. Both reconstructed, t-x domain and f-k domain in Figure 3.9f and Figure 3.9h demonstrates the limitation of the algorithm.

### **3.7.2 Uniform decimation for dipping events**

In order to generalize the algorithm for the interpolation, testing will be carried out with the uniformly decimation operators. Parametric reconstruction technique seems not to perform very well, when implemented on the uniformly decimated seismic section. In case of uniform decimation, replicas of events are created in the Fourier domain which is difficult to separate. But with the band-limiting approach like least square NFFT,

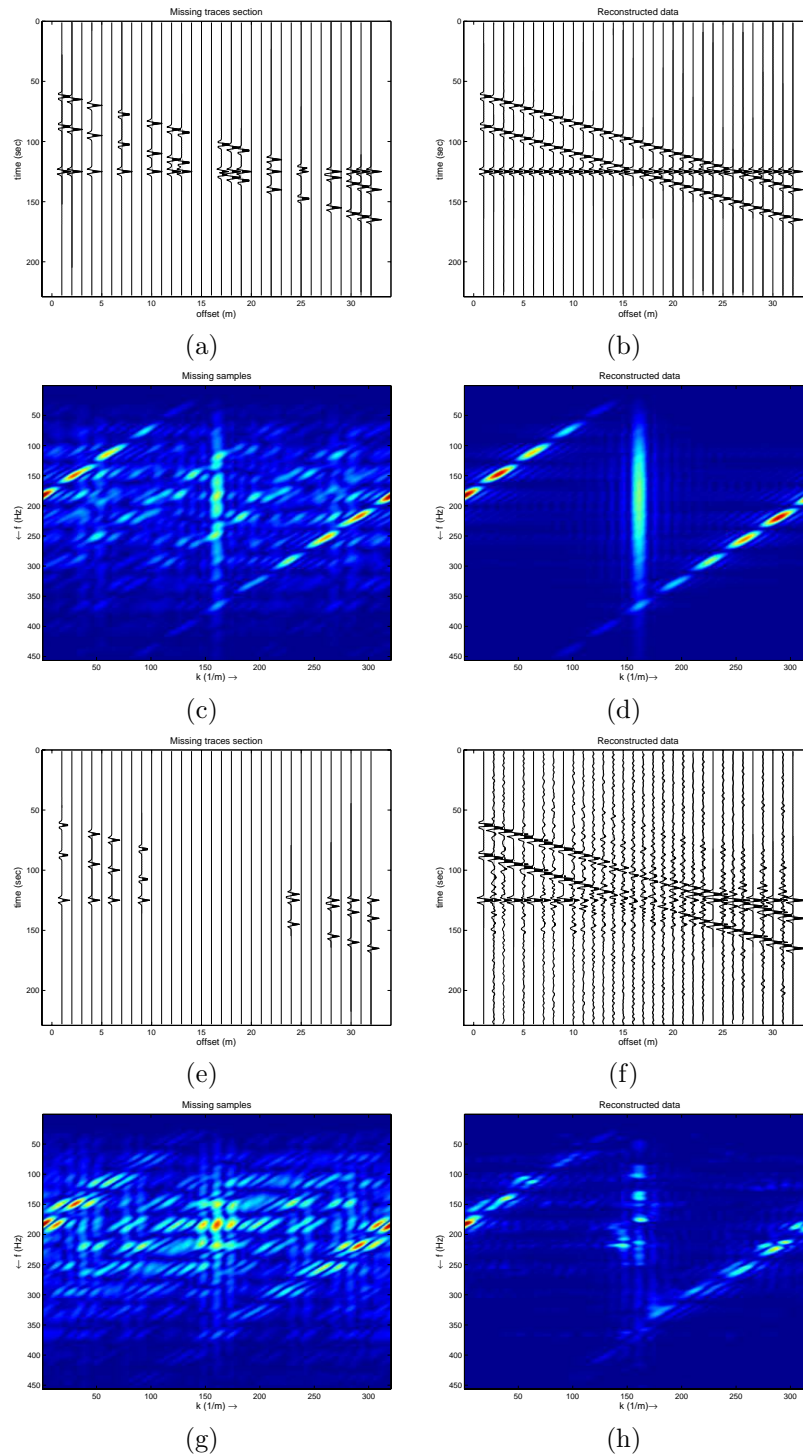


Figure 3.9: Reconstruction of random sampled seismic data. a) 50 % Decimated data. b) Reconstructed data for 50 % decimated data. c) Fourier domain for 50 % decimated data. d) Fourier domain for reconstructed data with 50% decimation . e) 80 % Decimated data. f) Reconstructed data for 50 % decimated data. g) Fourier domain for 80 % decimated data. h) Fourier domain for reconstructed data with 80% decimation.

replicated spectrum of event can be isolated in the low frequency of data. Its because of the higher power spectrum at low frequencies. Uniform decimation factors of 2 in Figure 3.10a and of 4 in Figure 3.10e are implemented.

2D Synthetic section is decimated by a factor 2 in Figure 3.10a. Exact replicas of planar and dipping events are created in FK domain of Figure 3.10c. Reconstructed data in Figure 3.10b and its Fourier domain in Figure 3.10d is recovered. On increasing the decimation factor to 4 in Figure 3.10e, have more replicas of planar and dipping events in Figure 3.10h as compare to 3.10d. But the recovered data in Figure 3.10f has well define events like in Figure 3.10f, setting reputation of algorithm to work on uniformly decimated data as well. Further for uniform sampling, like random sampling there is need of minimum number of samples so that algorithm can recover the data.

### **3.7.3 Hyperbolic events**

In case of hyperbolic events in Figure 3.11, data can always be windowed thus assuming that events are linear. But, already seen the application of least square NFFT on linear events. Applying LS-NFFT on the decimated data without windowing in Figure 3.11a. In upper part of reconstructed data in Figure 3.11b apexes are successfully reconstructed. But still some high amplitude noise is observed.

## **3.8 Conclusion**

Low computational cost of least square NFFT make it a robust and practical algorithm. This method successfully reconstruct the missing samples. This algorithm is effective both in case of random sampled data as well as uniform sampling. Algorithm can be easily extended to higher dimensions, and it will prove to be cost effective even for it. Though it is able to reconstruct the curved events. But a good windowing strategy which enforces linearity for curved events will sure provide better results in that case. NFFT

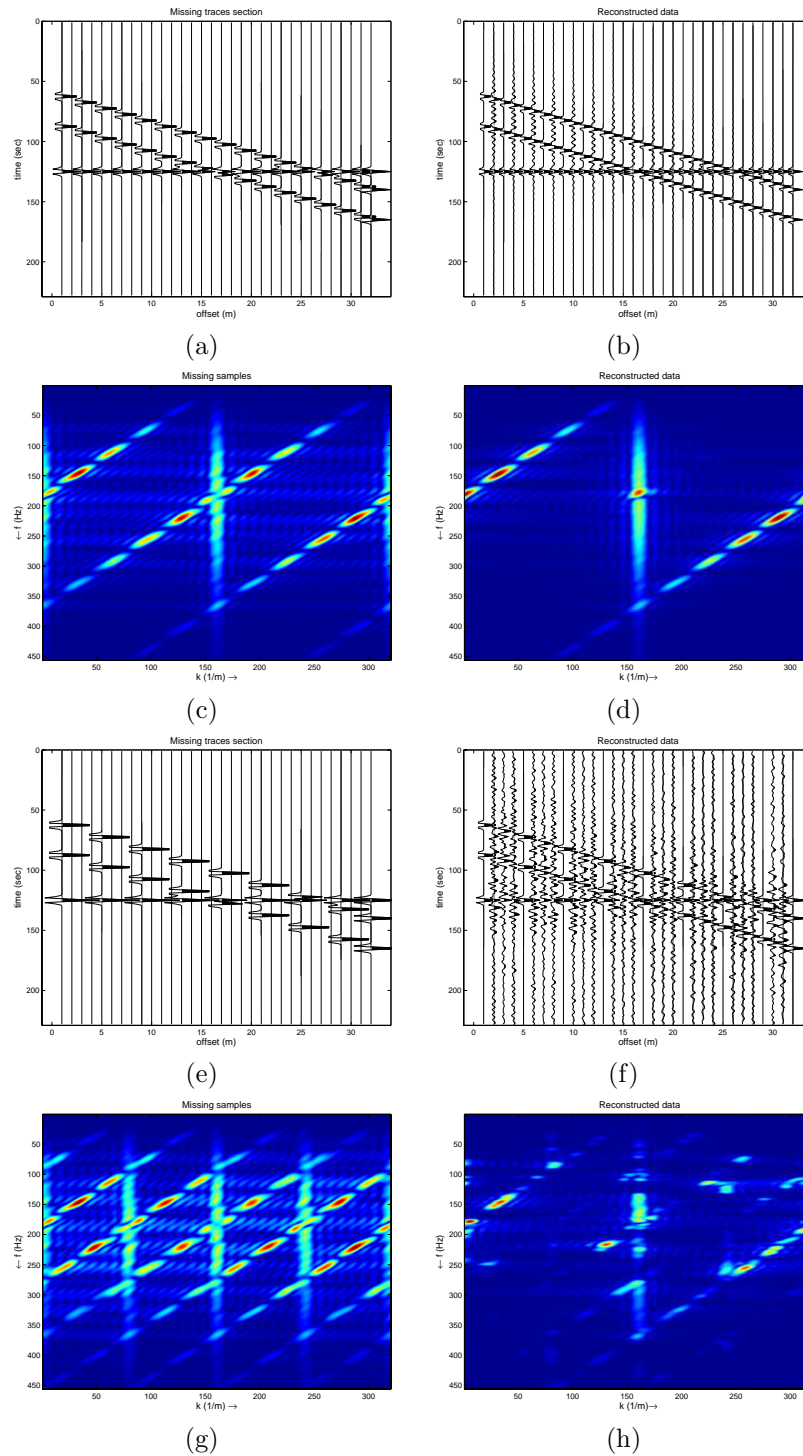
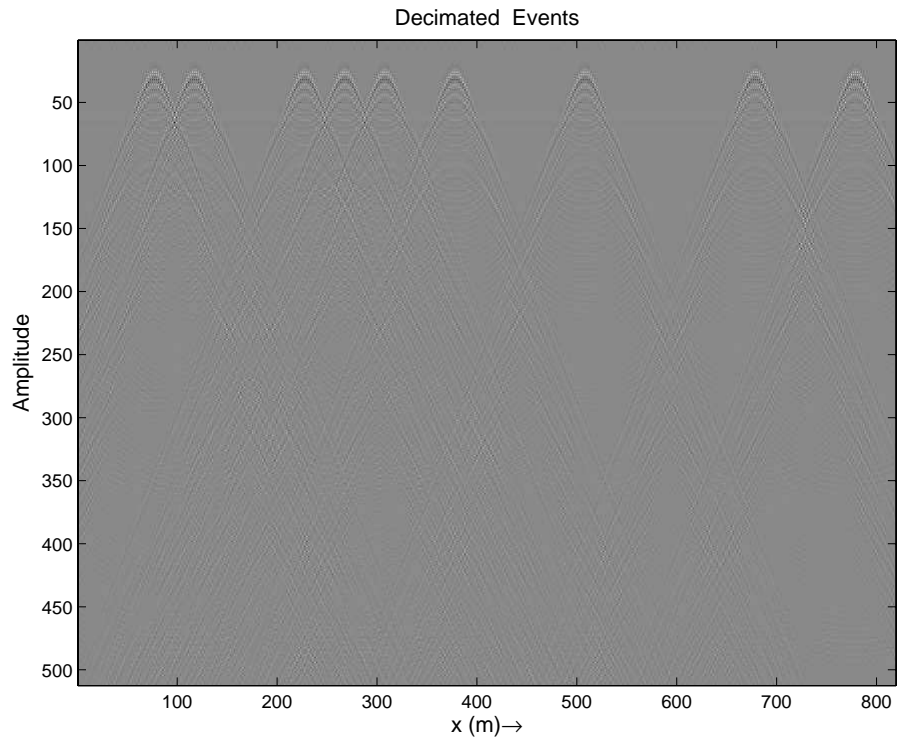
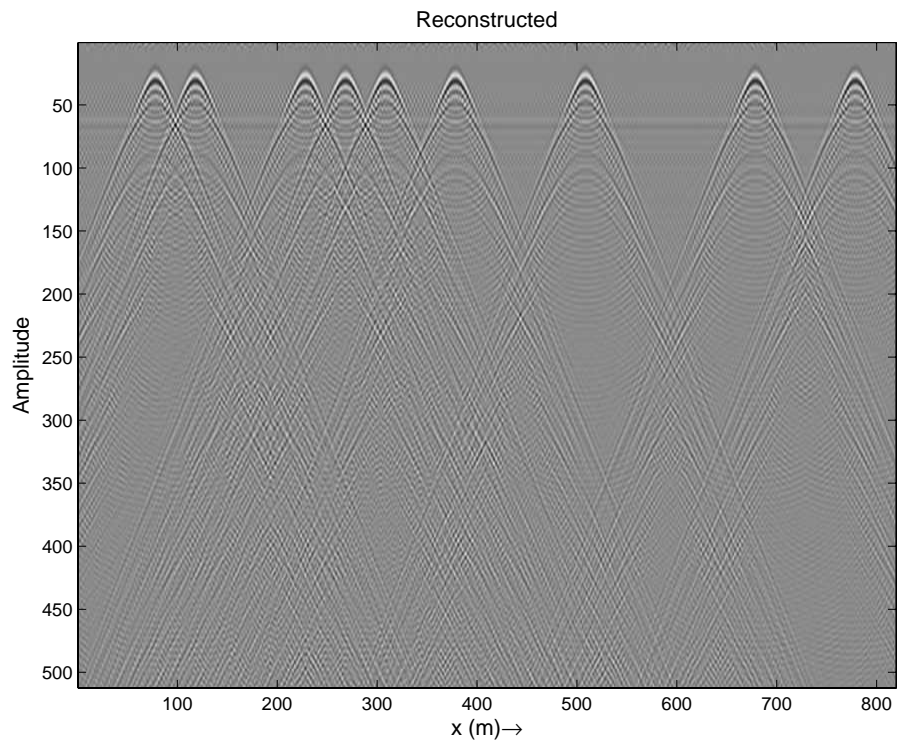


Figure 3.10: Reconstruction of uniformly sampled seismic data. a) Decimation by factor of 2. b) Reconstructed data. c) Fourier domain for decimation by factor of 2. d) Reconstructed data in Fourier domain for factor of 2 decimation. e) Decimation by factor of 4. f) Reconstructed data. g) Fourier domain for decimation by factor of 4. h) Reconstructed data in Fourier domain for factor of 4 decimation.



(a)



(b)

Figure 3.11: Reconstruction of hyperbolic events. a) Hyperbolic events with 20 % uniform decimation. b) Reconstructed hyperbolic events for 20 % decimated data.

and adjoint NFFT is a strong tool and can be used as an effective tool in other seismic processing steps.



## Chapter 4

# A simple algorithm for the restoration of clipped GPR amplitudes

### 4.1 summary

It is common in Ground Penetrating Radar (GPR) imagery to have missing or corrupted traces. This can be either due to obstacles, noise, technical problems or economic considerations. Antenna-ground coupling is another reason for clipped amplitudes in GPR data. Most commercially available software use the famous "rubber band interpolation", which uses the spline polynomial to undo the clippings. This method is a simple polynomial based interpolation which performs declipping without considering any prior knowledge about the signal.

In this chapter, a modified Projection on convex set (POCS) method is adopted for reconstruction of clipped amplitudes. Restoration of bandlimited GPR data which has undergone amplitude clipping is studied. This algorithm is tested on real GPR data which is clipped. To study the effectiveness of the technique, results obtained are compared with industry standard rubber band interpolation.

### 4.2 Introduction

Ground Penetrating Radar (GPR) methods are based on the same principle as seismic reflection methods. It is now a widely accepted geophysical technique. It is a non intrusive technique for detecting buried objects. The basic principle behind the GPR method is the transmission of electromagnetic energy into the earth and subsequent reflection from

the interfaces of differing dielectric. The GPR transmissions for the targeted subsurface form a synthetic aperture, whose impulse response is a spatially variant curve in the space-time domain. A common set up for GPR deploys a transmitter and receiver over a targeted zone. In some applications, trans-illumination of the volume under investigation is more useful. An example of GPR response is shown in Figure 4.1.

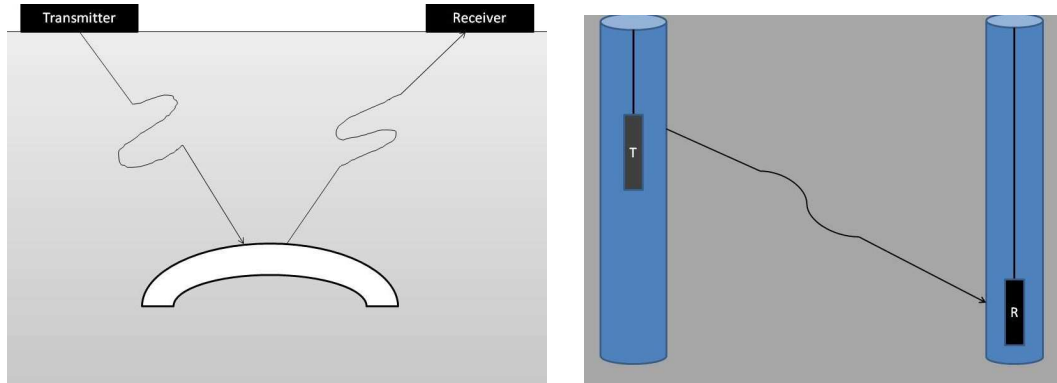


Figure 4.1: Ground penetrating radar (GPR) uses radio waves to probe the subsurface of lossy dielectric materials. Two modes of measurement are common. In the first, reflected or scattered energy is detected. In the second, effects on energy transmitted through the material are observed.

The response from the subsurface is produced from the combination of all buried units within the medium. This can be inverted using number of algorithms like Synthetic Aperture Radar (SAR) image formation techniques (Gazdag, 1978) and time domain standard back projection (Feng and Sato, 2004). These algorithms require a fine grid for spatial sampling and Nyquist-rate times samples of the received signals. Hence, the data acquisition for GPR is the bottleneck of the general subsurface imaging process.

In difficult terrain, due to manual error or some technical irregularity it is possible to have missing and corrupted traces in the data. This can result in a distorted subsurface image. In the case of GPR acquisition, the GPR unit is fired at regular time intervals and data will be collected in continuous trigger mode. There is no spatial direct measurement, so instead the operator tries to maintain a constant towing speed. Variation in towing

speed can not be ignored, and is evident from stretching of the GPR image, particularly at the end of the section.

If the signal is clipped, then the gap size can be large, so it can not be considered as a problem of interpolation of a regular spatial signal. The problem of interpolation of irregularly sampled signals is more complex and less well developed. It is mostly because of Shannon sampling theorem (Unser, 2000), which tells us,

”If a function  $f(x)$  contains no frequency higher than a peak frequency  $f_o$ , then it is completely determined by giving its ordinates as a series of points spaced  $T = \frac{2\pi}{f_o}$  seconds apart.”

This restricts the extension of Shannon sampling theory to signals defined over irregular grids. Still, with some constraints, algorithms have been proposed for reconstruction of band limited (Feichtinger et al., 1995b; Duijndam et al., 1999) and band unlimited signals (Naghizadeh and Sacchi, 2007a,b). These constraints limit the methods to certain applications. Large gaps in clipped GPR data is one restriction, and most band unlimited methods assume that the signal is stationary, which is generally not the case for GPR traces. Interpolation for reconstruction of seismic data (Sacchi et al., 1998; Xu et al., 2005; Liu and Sacchi, 2004; Naghizadeh and Sacchi, 2009b, 2008a) is performed along the lateral co-ordinate as field generally have irregularity of trace coverage. Here assumption that events are stationary is found to be effective.

In particular, the effectiveness of this theory and of the corresponding algorithms is restricted in the case of disparity compensated view interpolation; the derived constraints on the maximum gaps of irregular signals under perfect reconstruction conditions cannot be satisfied by irregular samples having big gaps. This restraint makes spline interpolation the only effective technique for the reconstruction in the GPR processing industry.

Other reconstruction algorithms have been proposed such as Projection on Convex Sets method (POCS) (Gerchberg and Saxton, 1972). This chapter uses one of the hybrid

method from the above categorized methods. Projection on convex sets along with non uniform fast Fourier kernel (Kunis and Potts, 2005) is used for solving the GPR clipping problem. This hybrid method will improve convergence rate and reduce the final reconstruction error. The main objective of this algorithm is to use oversampled gridding kernel, with POCS for reconstructing big gaps.

#### 4.2.1 Theory

POCS method is widely used for image reconstruction. The methodology involves finding a solution as an intersection property of sets rather than by minimization of a cost function. All image constraints are represented in a Hilbert space as a series of closed convex sets  $\{C_i | i = 1, 2, \dots, m\}$ , then each projection is done iteratively on the intersection. In simple terms, this algorithm estimates the missing data in a Hilbert space from its known parameters.

For example, if original signal has  $n$  properties  $S$ , then each property define one of the convex sets  $C_i$ . Also, the original signal will be part of all sets as well as of the intersection of sets as in Figure 4.2.

$$S \in C = \bigcap_{i=1}^n C_i \quad (4.1)$$

Equation 4.1 defines  $n$  sets for  $n$  properties of signal. The initial value of the signal is projected iteratively onto the intersection of all convex sets under the projection operator  $P$ . The optimal solution will be the point lying on the boundary of the intersection. Given the projection operator  $P_i$  onto  $C_i$ ,

$$S_{t+1} = P_n P_{n-1} \dots P_1 S_t \quad t = 1, 2, \dots \quad (4.2)$$

Equation 4.2 shows an iterative procedure for the signal with its projection operators. In Equation 4.2  $S$  converges to its limiting point of the intersection  $C$  in the Hilbert space

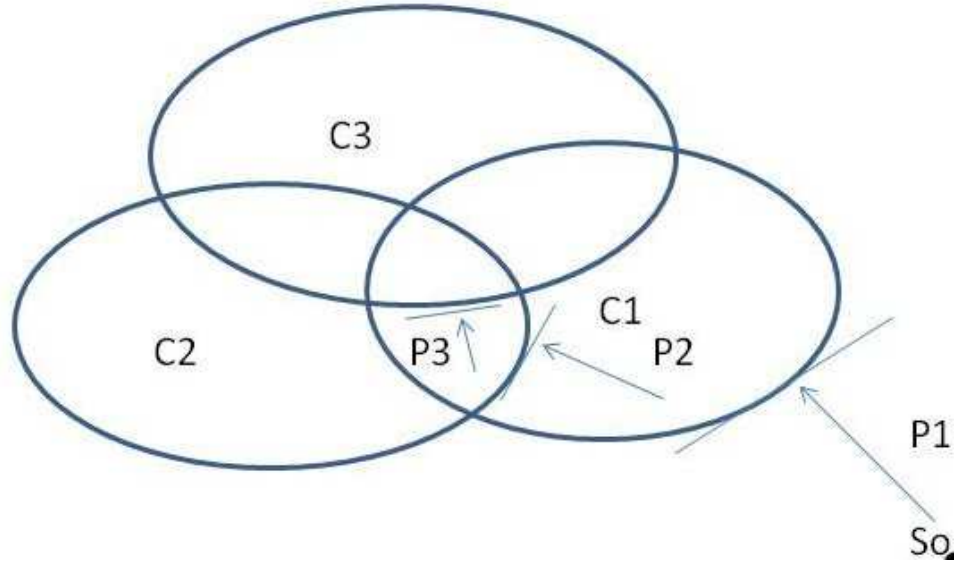


Figure 4.2: The principle of POCS Method

$H$ . The projection operator  $P_i$  will satisfy,

$$\|S - P_i S_i\| = \min_{k \in C_i} \|S - k\|, \quad (4.3)$$

where  $\|\cdot\|$  denotes the norm in  $H$ . The limiting point can be reached from  $n$  properties of  $S$  by using Equation 4.2.

The first step is to grid the data onto a regular grid using a gridding kernel. In this case the gridding kernel used is the kaiser Bessel kernel (Keiner et al., 2008), which convolves with the irregular data and distribute the samples onto a regular grid. This gridding kernel, which is used as non uniform fast Fourier transform (NFFT) kernel (Kunis and Potts, 2005), acts as simple FFT when the samples are already on a regular grid. The point to be stress that, if the gaps are small then the simple FFT kernel can be applied instead of NFFT.

POCS is iterative and typically projects consecutive solutions onto consecutive properties sets. Each iteration is followed by the NFFT kernel, which is the FFT when sampling is regular enough. A threshold is applied to the Fourier domain leaving components greater than the threshold as zero. During the first few iterations, sample points

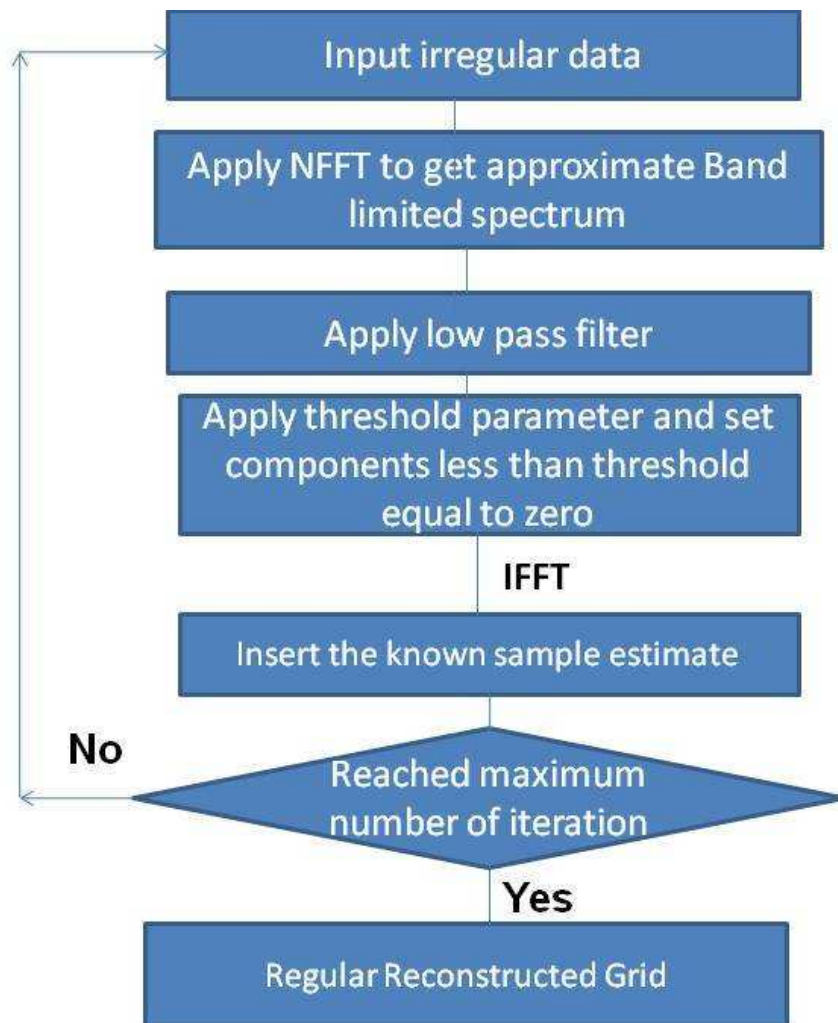


Figure 4.3: flowchart

with high energy are restored. In each iterations, higher frequencies are made zero in the frequency domain. The threshold parameter enforces a cut off in amplitude which gives some amplitude to unknown values. After this, the value of known components are restored by replacing them with their true values. This will reconstruct the high frequency values. Samples will be reconstructed in each iteration (Figure ??). The whole process can be written in form of Equation 4.4.

$$S_k = S_{obs} + (I - S)F^{-1}T^k B(NFFT)S^{k-1} \quad (4.4)$$

where,  $S_{obs}$  is a original data at kth iteration.  $S_{obs}$  will keep getting updated until it finally converges to a solution. NFFT and  $F^{-1}$  represents non uniform fast Fourier transform and inverse fast Fourier transform which operates on  $t$ .  $S$  is a sampling operator that identifies known and unknown values.  $T^k$  is threshold operator with elements.

$$T^k = \begin{cases} 0, & F_{k-1} \geq l_k \\ 1, & F_{k-1} < l_k \end{cases} \quad (4.5)$$

Where,  $F_{k-1}$  denotes the Fourier domain representation of the reconstructed signal after the  $(k - 1)$ th iteration.  $l$  represents the  $N$  dimensional threshold set  $l = l_1, l_2, \dots, l_N$  where  $l_1 > l_2 > l_N$  and  $N$  denotes the maximum number of iterations.

## 4.3 Experiment

### 4.3.1 Synthetic data

Figure 4.4a shows three a simple section with two planar event and a dipping event. Figure 4.4b indicates the clipped version of figure 4.4a. Scale shows that original section has maximum amplitude of 3 where as figure 4.4b is clipped to 0.7. 4.4c shows the reconstructed amplitude which is approximately similar to figure 4.4a. Figure 4.5a is an

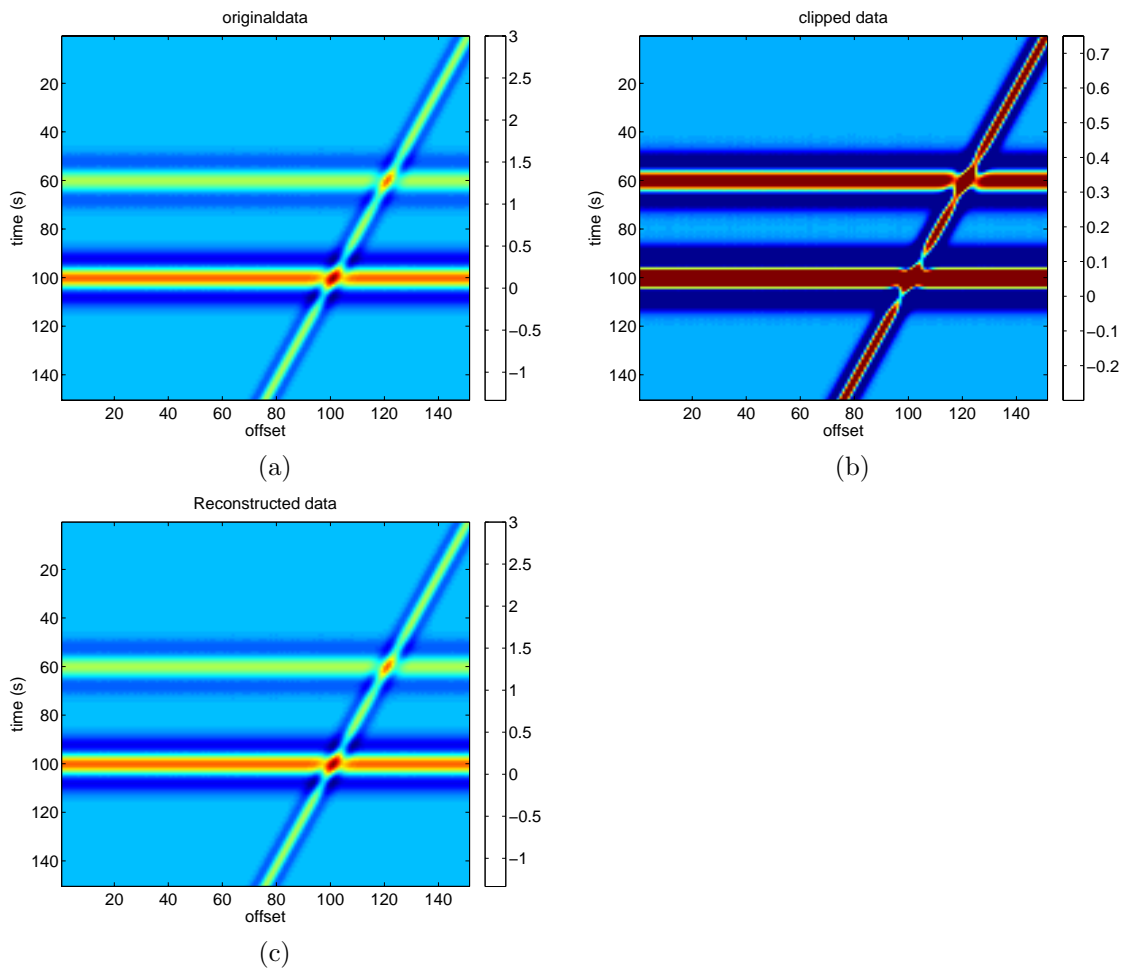


Figure 4.4: Synthetic sections. a) Original synthetic data. b) Clipped data. c) Reconstructed data.



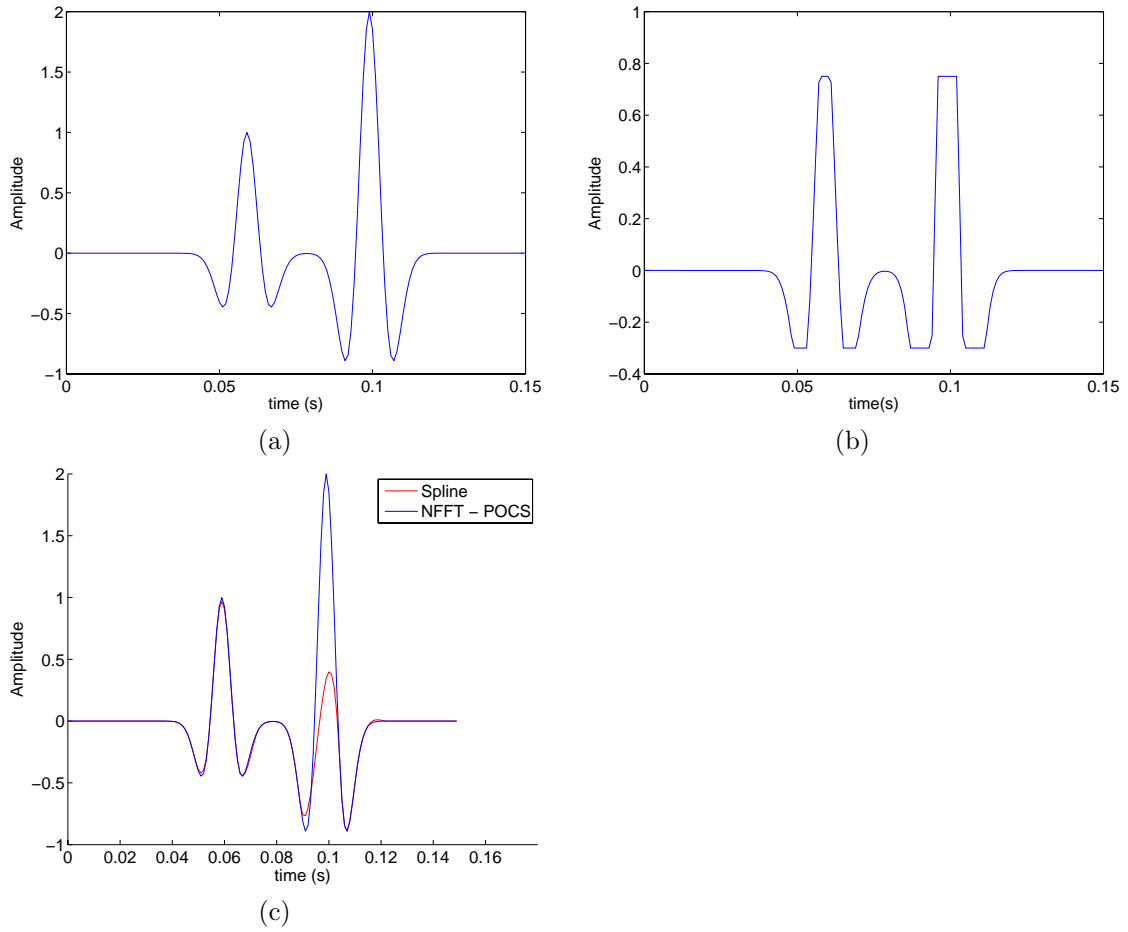


Figure 4.5: a) Synthetic trace. b) Clipped trace. c) Reconstructed trace

another synthetic trace made up of two ricker wavelets. To test the algorithm data is being clipped in Figure 4.5b. Followed by reconstruction using conventional spline and NFFT POCS. NFFT POCS is able to restore the clipping perfectly where as spline gave completely inaccurate reconstruction of second peak in Figure 4.5c.

### 4.3.2 Real data

The proposed reconstruction algorithm has been tested experimentally on the GPR data. The data is acquired along an ice sheet. Snow and ice are ideal GPR media because their stratification presents reflecting horizons with great continuity and interesting configurations. Due to antenna-ground coupling or some other technical issue the amplitudes in the acquired GPR data are clipped. They are clipped below  $-2^{15}$  and above  $2^{15} - 1$ , so there is no positive amplitude greater than 32767, and no negative amplitude below -32768.

Figure 4.6 shows the GPR data with a lake bottom reflection. The sampling rate for the acquired data set is 4ns. The bandwidth of data is 0-1250 MHz. Clipped amplitudes are clearly visible in Figure 4.6 along the top horizon. The data is comprised of 300 traces, each sampled for 750 ns. Sample traces from the data set can be seen in Figure 4.7, where it is clearly evident that it is clipped with maximum amplitude of 32767 and minimum amplitude of -32768.

The first step in commercial GPR processing software is the implementation of rubber band interpolation (spline interpolation). Figure 4.8 shows the effectiveness of spline interpolation for a GPR trace. Implementation of different interpolation technique on GPR data is carried out. Cubic interpolation along with linear and nearest neighbour interpolation is compared with spline interpolation. Cubic spline interpolation is implemented on the trace, it is a piecewise third-order polynomial passing through set of points. The result is not very effective with any other method except spline interpolation.

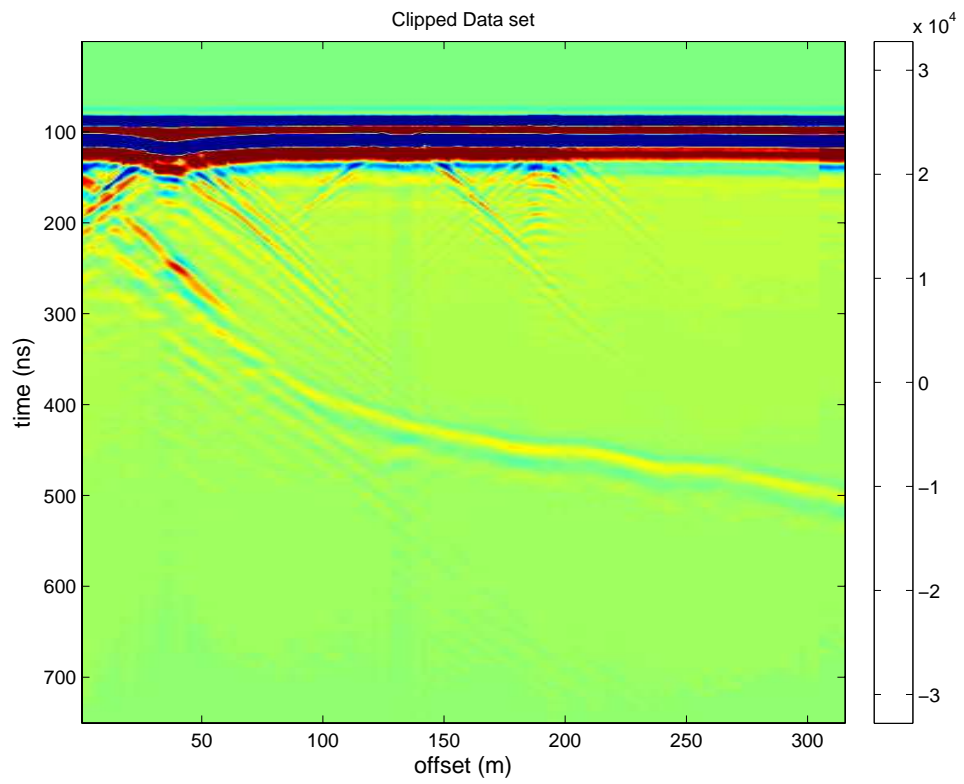


Figure 4.6: Acquired GPR data with clipped horizons.

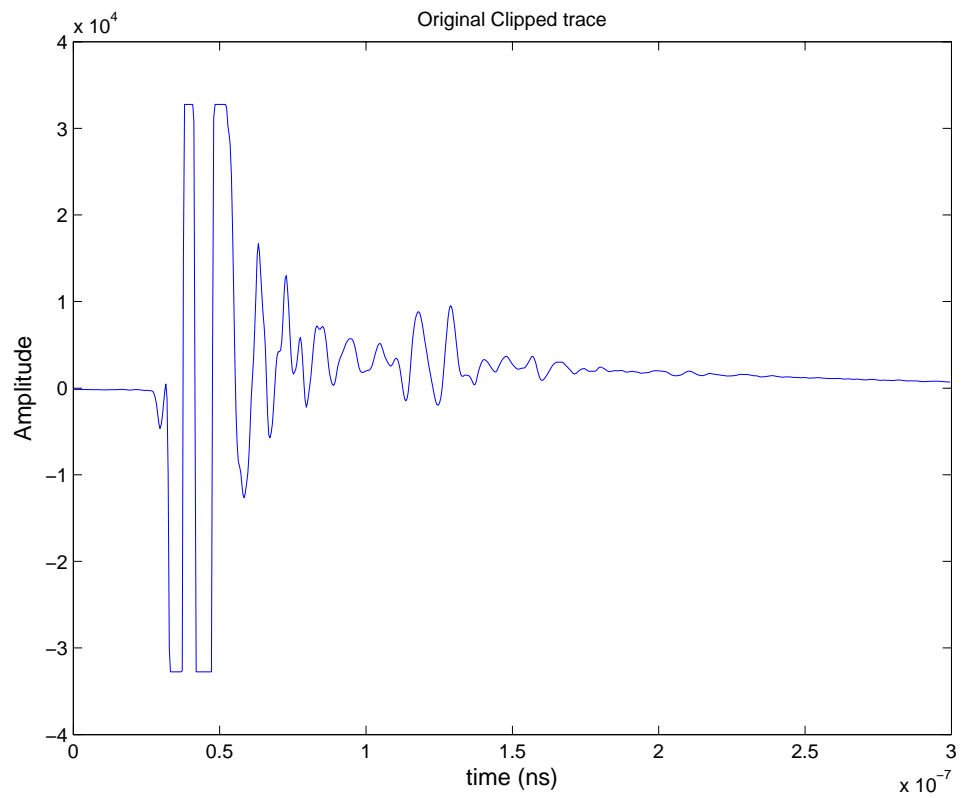


Figure 4.7: Random extracted clipped trace from acquired GPR data.

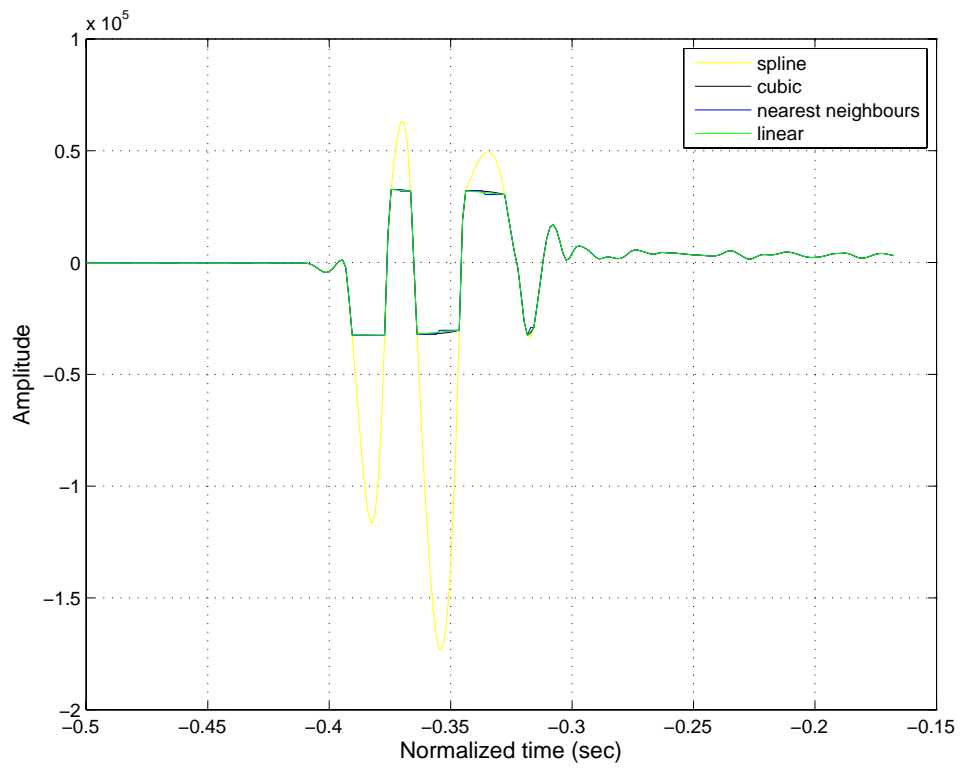


Figure 4.8: Comparative study for reconstruction of clipped amplitude

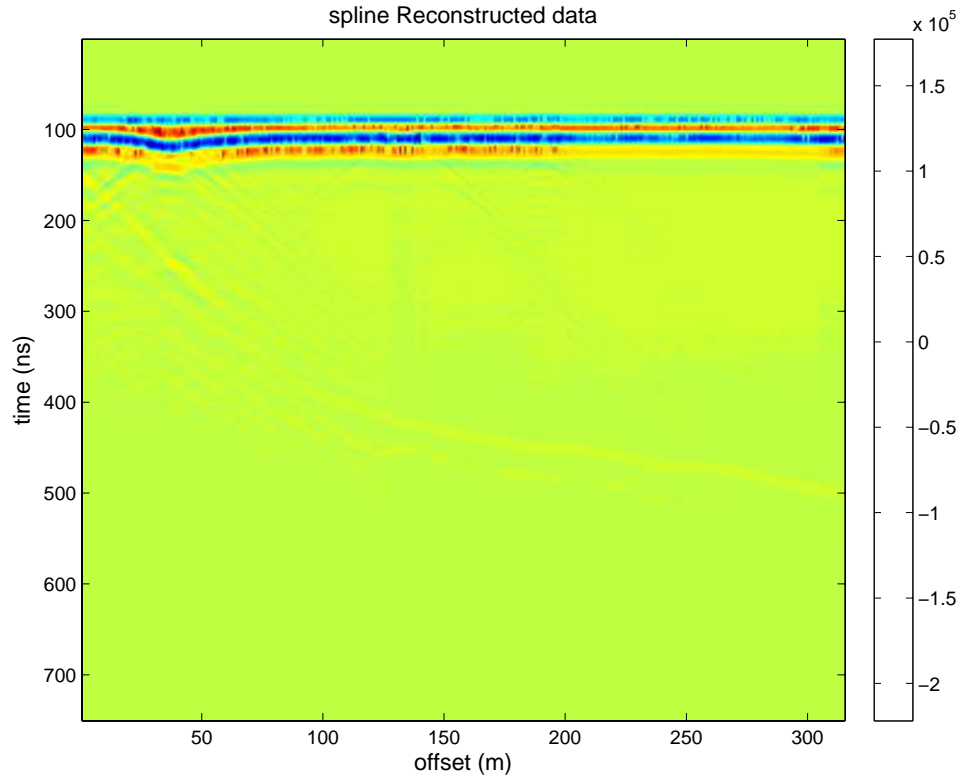


Figure 4.9: Spline interpolation of clipped GPR data.

This effectiveness of spline interpolation makes it a favourite module in declipping or desaturation for various GPR processing software. However, this method does not use any information from the signal. The whole reconstruction procedure is based on the fact that the interpolater is calculated on the basis of a few nearby data points. Figure 4.8, in which spline is use for reconstruction of single trace from GPR data, shows relatively good results as compared to other interpolators. Amplitudes are restored in Figure 4.9 on application of the spline based interpolation.

Effectiveness of modified POCS method is validated by comparing it with spline interpolation on different GPR traces. Figure 4.10a shows that for the 10th trace, there is difference in the reconstructed negative and positive clipping as compared to the spline method. Figure 4.10b shows a difference in all four reconstructed clippings, these four clips represent four different horizons. All of the clippings reconstructed have a significant

error in both negative and positive amplitudes. Figure 4.10b shows another trace, where spline and Modified POCS give us similar results. Figure 4.10c and 4.11a also shows the difference in reconstructed amplitudes.

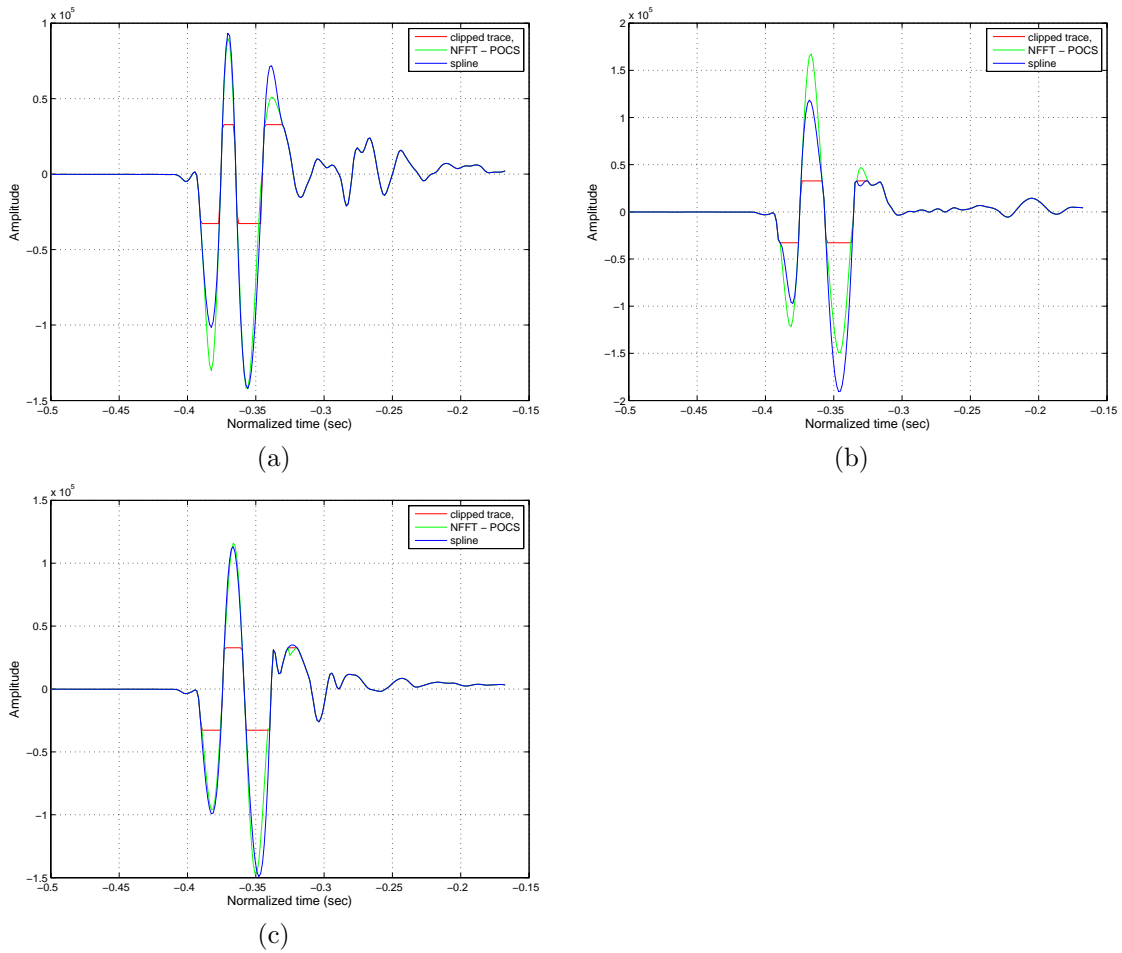


Figure 4.10: Restored clipped amplitudes using Spline and Hybrid NFFT POCS. a) 10th trace from restored GPR section. b) 30th trace from restored GPR section. c) 50th trace from restored GPR section.

Figure's 4.11 and 4.12 shows that it is not always the case that Modified POCS is constructing amplitudes smaller than the spline method. It can be seen in Figure 4.11a that the amplitude restored in the first clipping is higher compared to spline based technique. Also, in Figure 4.11c, the restored clips have higher amplitude than the splines. In Figure's 4.10, 4.11 and 4.12 results obtain from our method differs from conventional spline based method.

Figure 4.13 shows the reconstructed GPR section. Horizons are successfully reconstructed, and the energy is continuous along the horizon, whereas the spline based re-



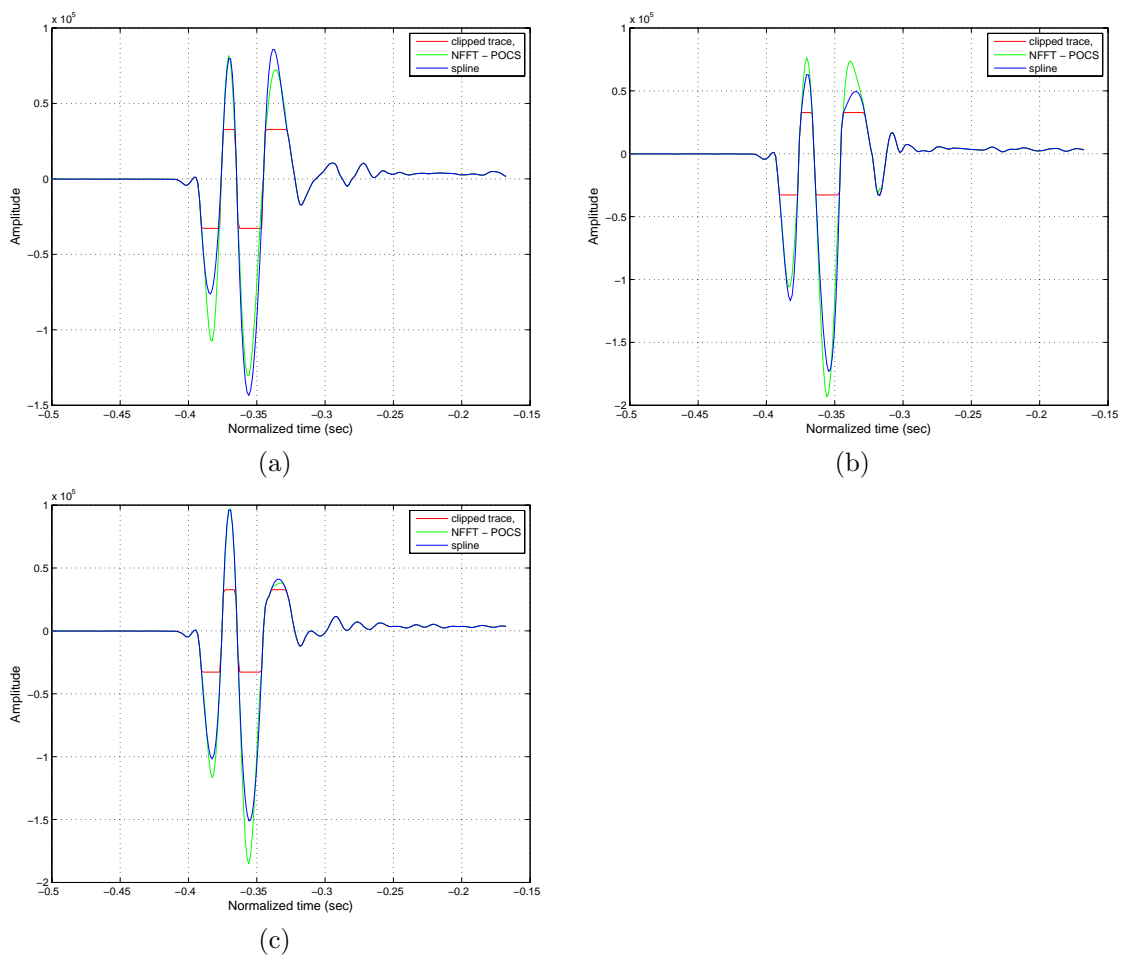


Figure 4.11: Restored clipped amplitudes using Spline and Hybrid NFFT POCS. a) 100th trace from restored GPR section. b) 150th trace from restored GPR section. c) 200th trace from restored GPR section.

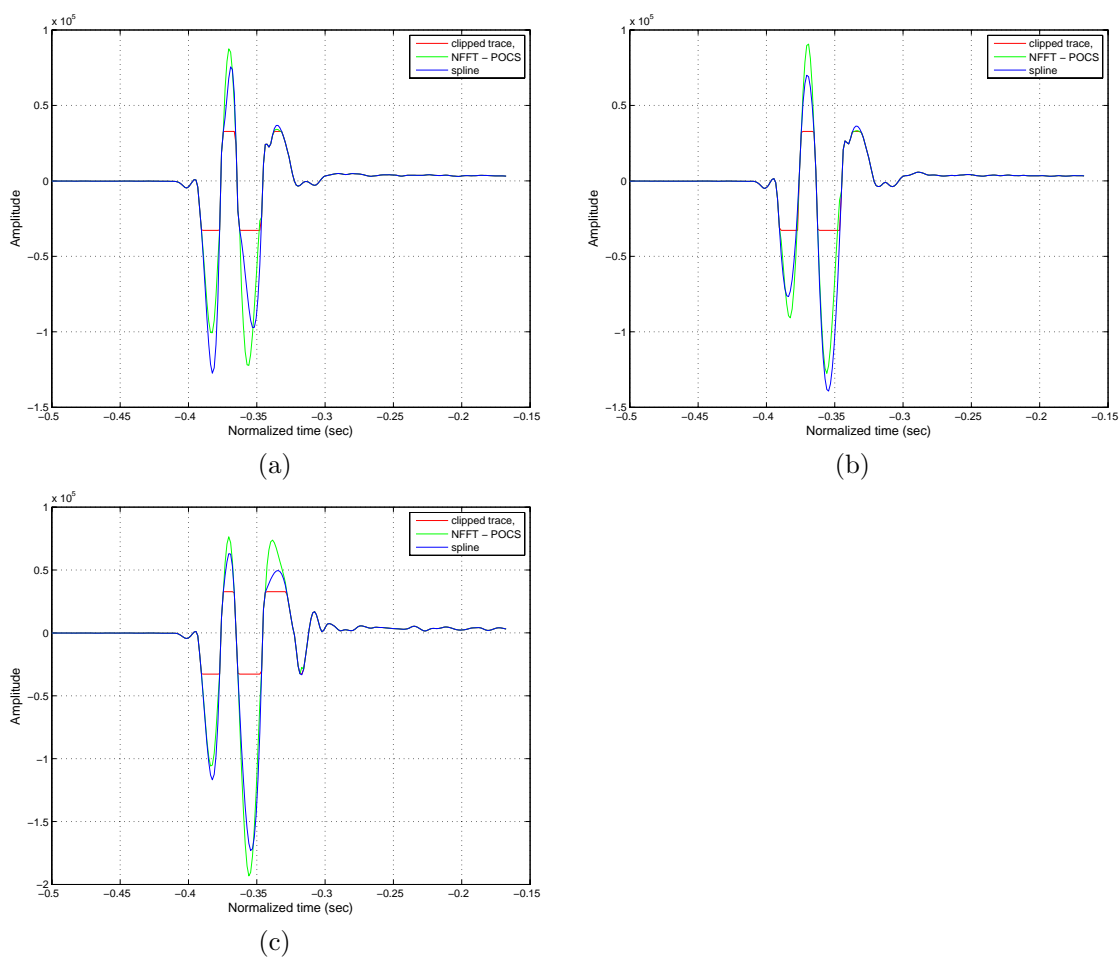


Figure 4.12: Restored clipped amplitudes using Spline and Hybrid NFFT POCS. a) 250th trace from restored GPR section. b) 280th trace from restored GPR section. c) 300th trace from restored GPR section.

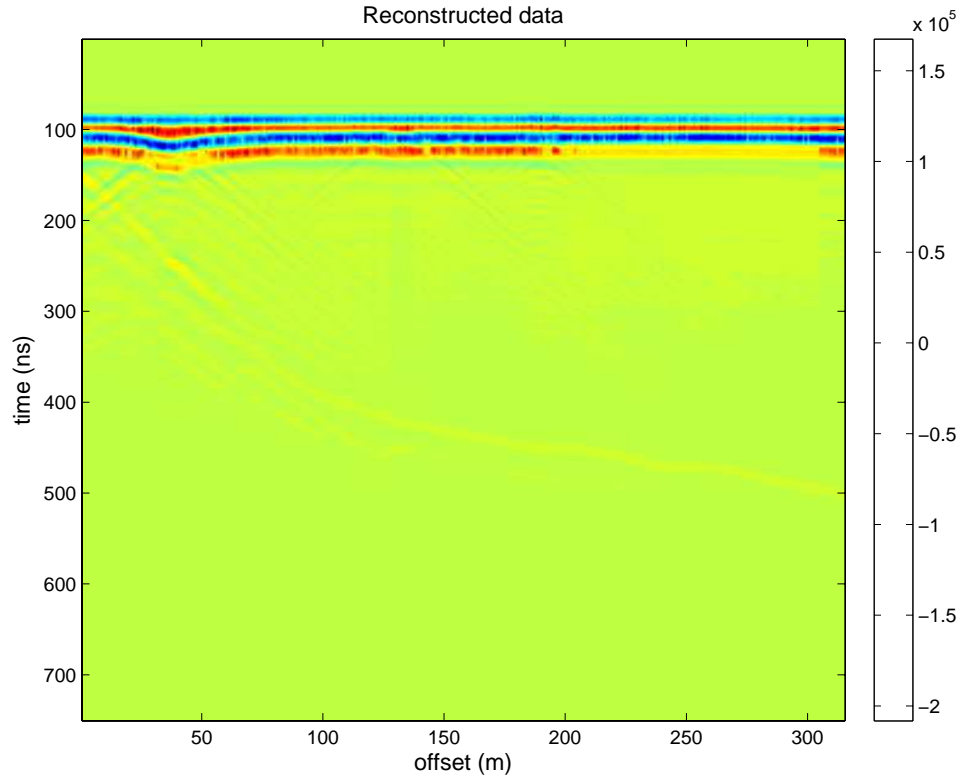


Figure 4.13: Reconstructed data using Hybrid NFFT-POCS

constructed GPR section in Figure 4.9 seems to lose energy in the first and second horizons. It is also noted in Figures 4.10, 4.11 and 4.12 that there is a difference in restored amplitudes from both methods.

Residuals between original clipped GPR data and POCS-based reconstruction are computed. Figure 4.14 shows smoothness in the restored amplitudes across the horizons. The first, second, and third horizon show constant lateral smoothness in reconstructed energy across the horizons in each layer. The residual of original clipped data and spline-based method in Figure 4.15 shows that energy is not constant in the top horizons, which indicates the drawback of spline interpolation. Figure 4.16 indicates the difference in the reconstructed horizons between Modified POCS and spline method, which reflects the difference in the reconstructed data.

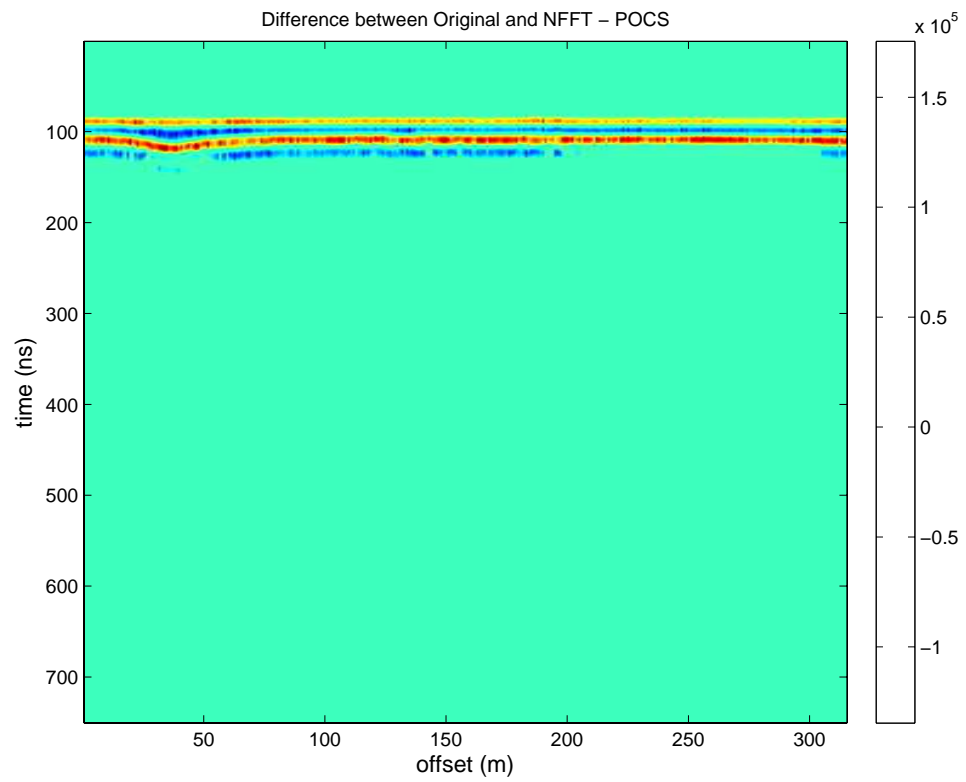


Figure 4.14: Residual between original and reconstructed section using Hybrid NFFT-POCS.

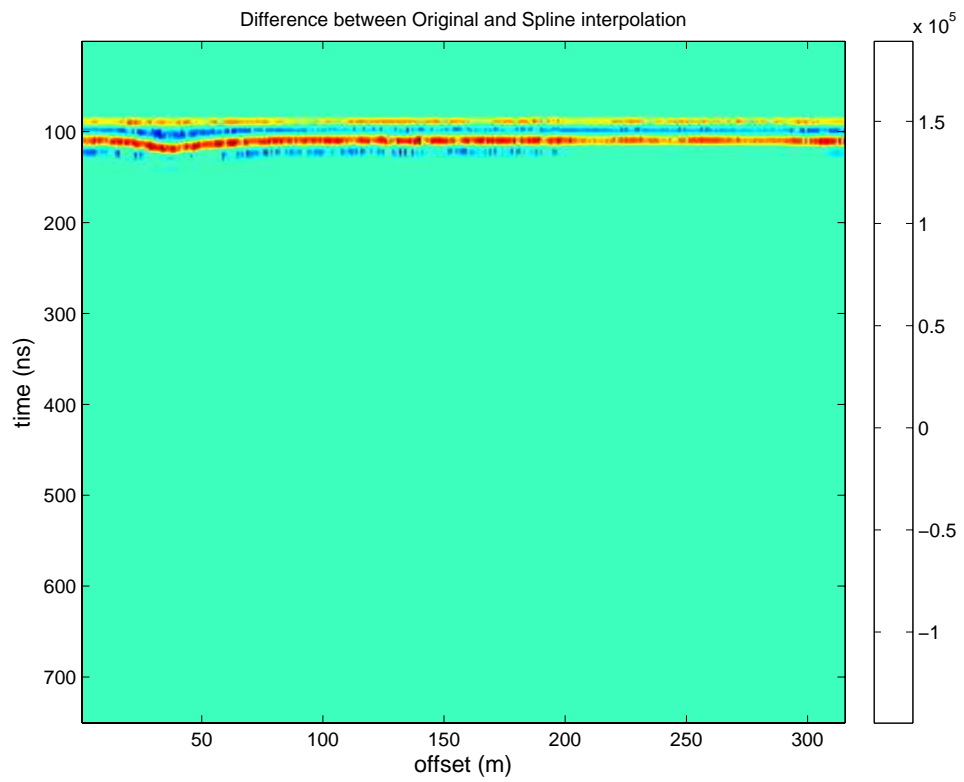


Figure 4.15: Residual between original and reconstructed section using spline interpolation.

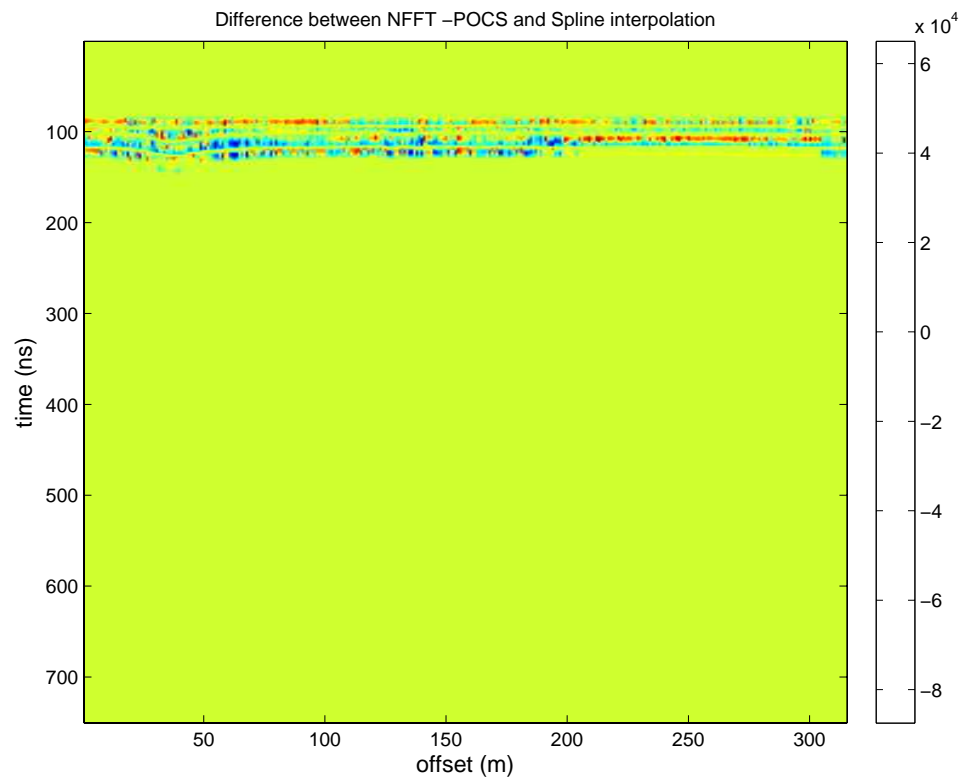


Figure 4.16: Residual between restored GPR section using spline and Hybrid NFFT-POCS.

## 4.4 Conclusion

In this chapter, an algorithm for clipped amplitude restoration using hybrid POCS has been presented and tested. It is able to completely restore the clipped amplitudes of GPR data. Two different methods for estimating the clipping have been tested. The first one is conventional method of spline interpolation, which is largely adopted in GPR industry. The second is hybrid POCS, which uses *a priori* information from the signal to recover clipped amplitudes. A comparative study is done, which showed that Hybrid POCS is better than conventional spline interpolation. Hybrid POCS is better technique due to improved lateral continuity of the energy across the horizons in reconstructed data.

# Chapter 5

## Conclusion

This thesis studies the reconstruction of bandlimited seismic data using Fourier reconstruction techniques. The main objective was to illustrate the use of Fourier reconstruction techniques of seismic data and to introduce Fourier method that provide robust and efficient tools for seismic data reconstruction.

The main contributions of this thesis to the field of interpolation in geophysics can be summerized as follows:

- Analysis of the presently available slow methodology, known as Adaptive Conjugate Toeplitz (ACT) which uses simple DFT for reconstruction of missing samples in regular and irregular data. This methodology gives good interpolation result, but it is slow and inefficient (**chapter 2**).
- A novel reconstruction method called the Kaiser Bessel Non uniform Fast Fourier Transform (NFFT). As the name suggest Kaiser Bessel NFFT is a convolution of the Fast Fourier transform with a Kaiser Bessel filter. This methodology is tested for both regular and irregular decimations. It allows the replacement of slow DFTs with faster NFFTs (**chapter 3**).
- A problem of clipped amplitudes in GPR data has been solved. The problem of restoring clipped amplitudes cannot be solved using simple techniques, due to the presence of big gaps. A hybrid method is implemented which involves a combination of NFFT with POCS (Projection on Convex Set). The declipping algorithm was tested sucessfully with synthetic and real world examples. The results obtained through this proposed hybrid methodology are better than conventional methods



like spline interpolation (**chapter 4**).

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